

Tech Comparison

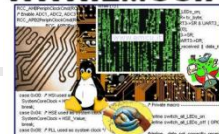
Modena, 15 Aprile 2014



Power 'n More

SILICA Power Solutions

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Breakthrough in Power Electronic

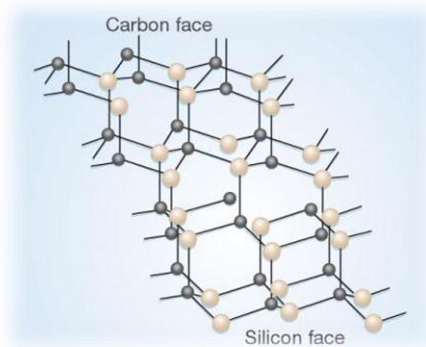
SiC Technology SiC MOSFETs & SiC Rectifiers



- Silicon Carbide (chemical formula SiC), is a compound of silicon and carbon
- Most Silicon Carbide is of synthetic nature



Crystal...



...made of silicon and carbon atoms



A
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Seals rings for industrial applications

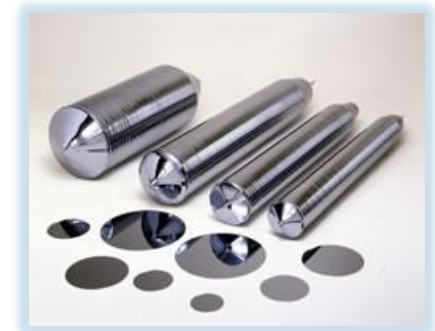


Cutting tools

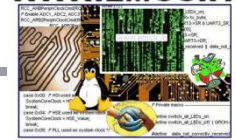


Sports car brakes

AND...



Semiconductor



Mechanical properties: Mohs Hardness scale

Mohs Hardness Scale		
Hardness Number	Original Scale	Modified Scale
1	Talc	Talc
2	Gypsum	Gypsum
3	Calcite	Calcite
4	Fluorite	Fluorite
5	Apatite	Apatite
6	Orthoclase	Orthoclase
7	Quartz	Vitreous Silica
8	Topaz	Quartz or Stellite
9	Corundum	Topaz
10	Diamond	Garnet
11	...	Fused Zirconia
12	...	Fused Alumina
13	...	Silicon Carbide
14	...	Boron Carbide
15	...	Diamond

Almost as hard as Diamond!!!

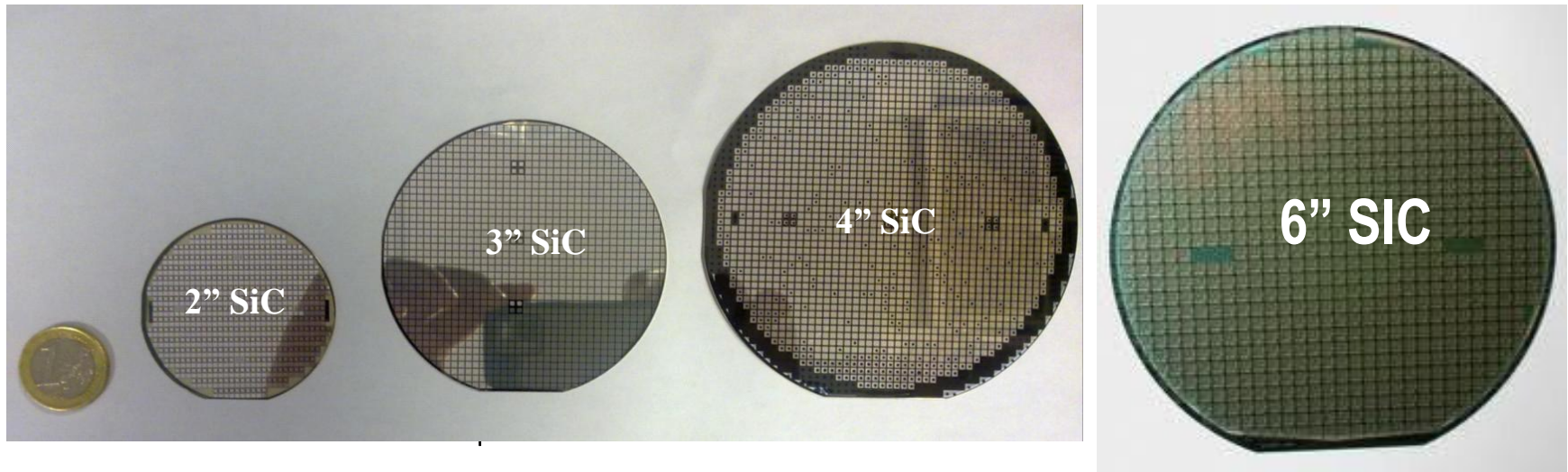


Electrical properties

Properties at 300 K, $1^{e15} - 1^{e16} \text{ cm}^{-3}$	E_G (eV)	E_{BR} (V/ μm)	Thermal conductivity (W/cm/K)
Material			
Si	1.12	20	1.5
4H-SiC	3.26	200	4.5
Diamond	3.45	560	20

Much wider band gap (E_G) than silicon resulting in a higher critical electrical field (E_{BR}).

ST has Well-established Expertise on SiC Material & Devices: Wafer Size Evolution



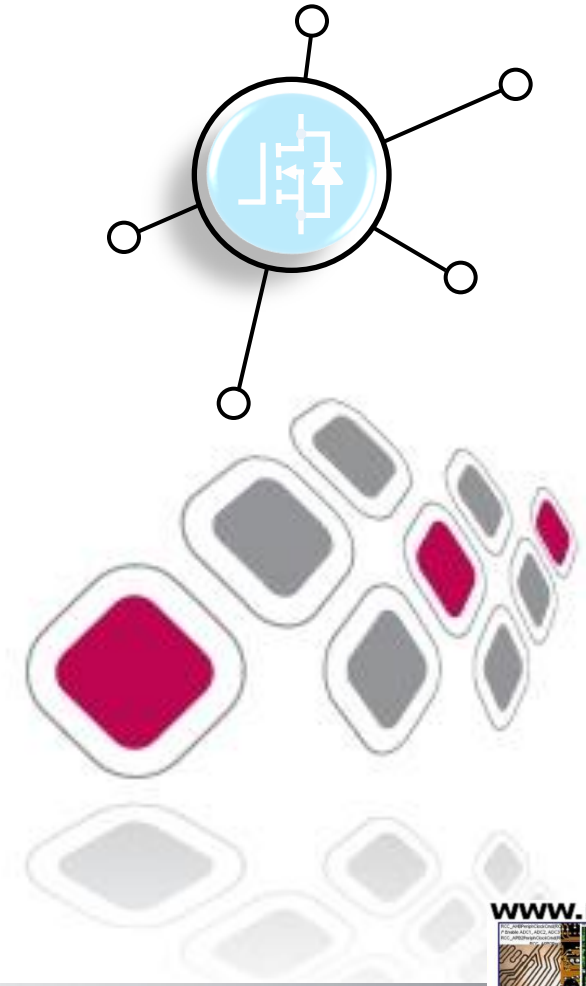
Both MOSFET and Diode in SiC are manufactured in the same fab

Silicon vs. Wide-bandgap Silicon Carbide					
Parameter	Symbol	SI	Silicon	4H-SiC	Benefits
Bandgap	E_g	eV	1.1	3.3	Lower Leakage, Higher T_j
Electron Saturation Velocity	V_s	cm/s	1×10^7	2×10^7	Higher working frequency
Electron Mobility	μ_n	$\text{cm}^2/\text{V}\cdot\text{s}$	1350	947	
Dielectric Constant	ϵ_r	-	11.8	9.7	
Critical Electric Field	E_c	V/cm	0.3×10^6	3×10^6	Lower On-Resistance
Thermal Conductivity	k	W/cm-K	1.5	4.5	Higher thermal handling capability

➤ **Wide-bandgap semiconductors are materials with a bandgap significantly larger than 1 eV.**

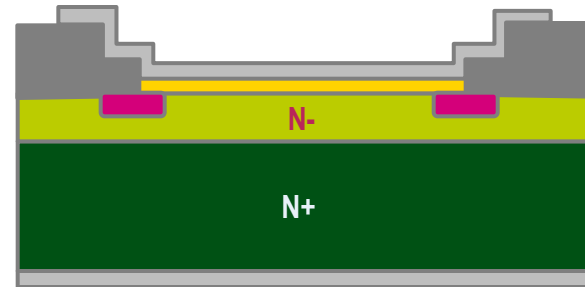
SiC Diode

Gen1 & Gen2

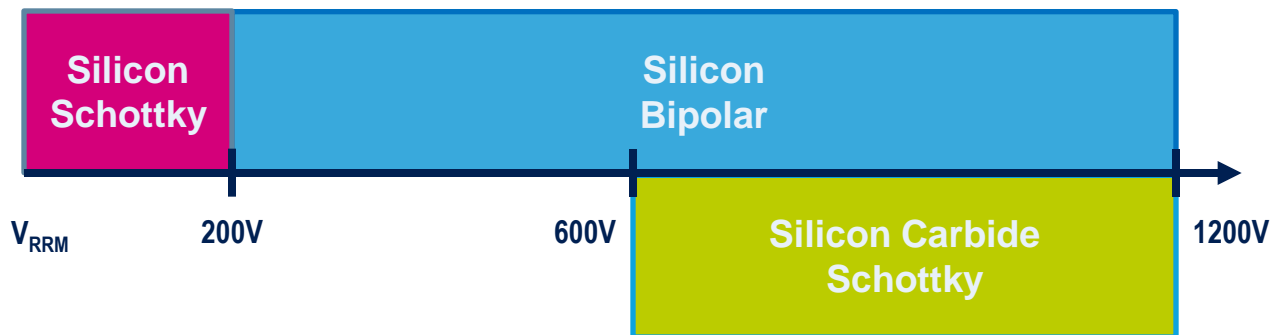


SiC Schottky Diode

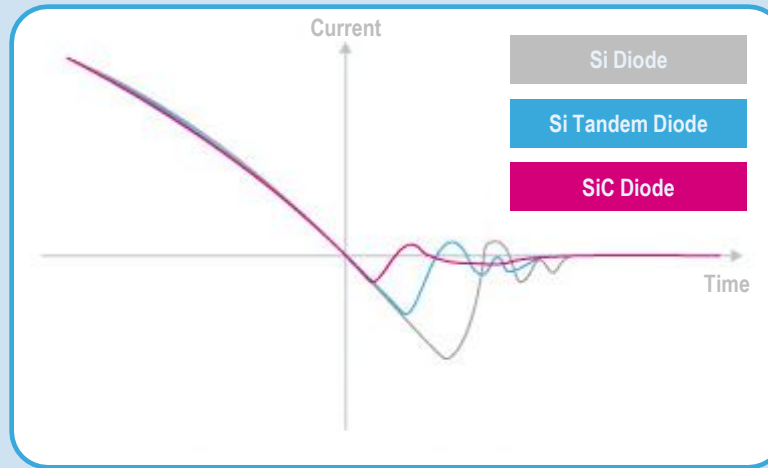
- 4 times better dynamic characteristic and 15% less forward voltage than silicon diodes
- Need less thickness and resistivity to sustain the same breakdown voltage
- Outstanding electrical characteristics:
 - SiC $E_{BR} = 200 \text{ V}/\mu\text{m}$
 - Si $E_{BR} = 20 \text{ V}/\mu\text{m}$
- Repetitive Peak Reverse Voltage (V_{RRM}) up to 1200V



Schottky Diode structure diagram

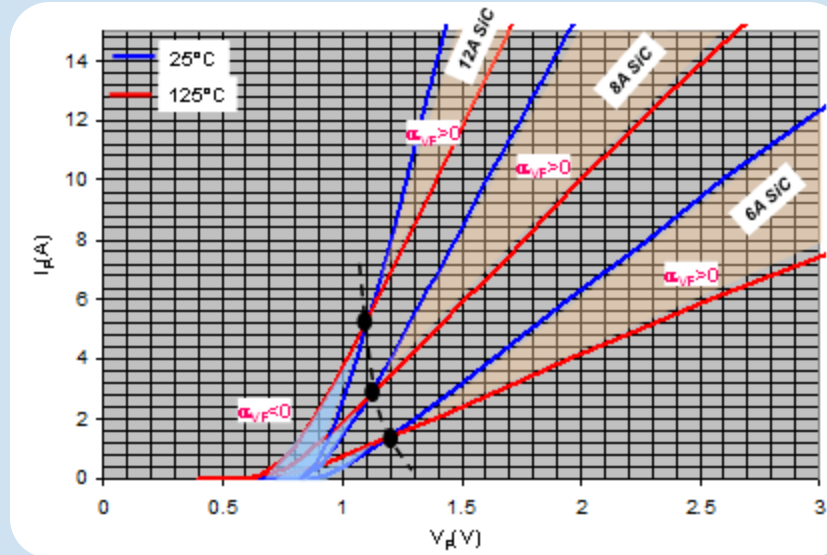


Ultra-fast Diode Recovery Time



HV Rectifiers	Si Diode	Si Tandem Diode	SiC Diode
Recovery Time	High	Low	Extremely Low
Switching losses	High	Low	Negligible
Reverse Recovery Charges Q_{rr} (nC)	150	50	6

Drawback of Diodes made out of SiC material

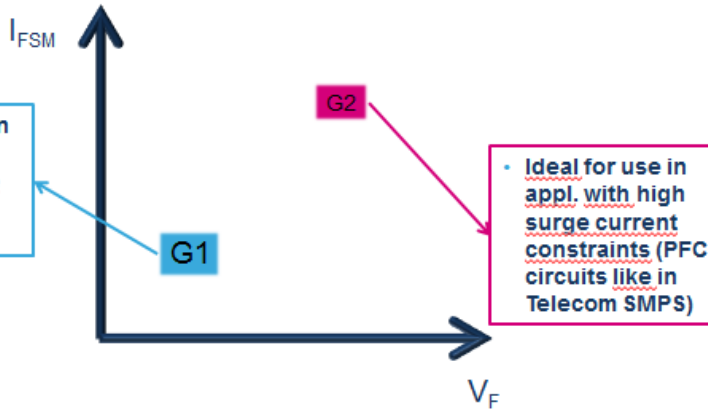


SiC material	Consequence	Drawback
SiC features a positive thermal coefficient	The higher the temperature, the higher the V_f	Thermal runaway risk when high current is applied!

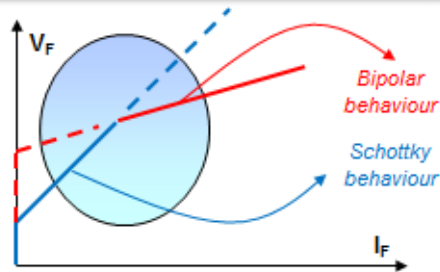
Gen1 or Gen2 for Designer's choice

12

Gen1 or Gen 2? Vf or IFSM trade off



The BEST out of 2 Technologies



The addition of P+ implantation in the schottky structure creates P/N junctions. The surge forward current capability can be increased while keeping $T_J < T_{J(MAX)}$

10 times better Surge Capability

Datasheet:

New voltage rating for higher reverse safety margin

Symbol	Gen 2	Parameter	Value	Unit
V_{RRM}		Repetitive peak reverse voltage	650	V
$I_{F(RMS)}$		Forward rms current	21	A
$I_{F(AV)}$		DPAK, $T_c = TBD$ °C, $\delta = 0.5$	6	A
		TO-220AC, $T_c = 122$ °C, $\delta = 0.5$		
		D ² PAK, $T_c = TBD$ °C, $\delta = 0.5$		
I_{FSM}		$t_p = 10$ ms sinusoidal, $T_c = 25$ °C	60	A
		$t_p = 10$ ms sinusoidal, $T_c = 125$ °C	52	A
		$t_p = 10$ μ s square, $T_c = 25$ °C	400	A
T_{stg}		Storage temperature range	-55 to +175	°C

Improved surge capability
 $I_{FSM} = 10 \times I_{F(AV)}$

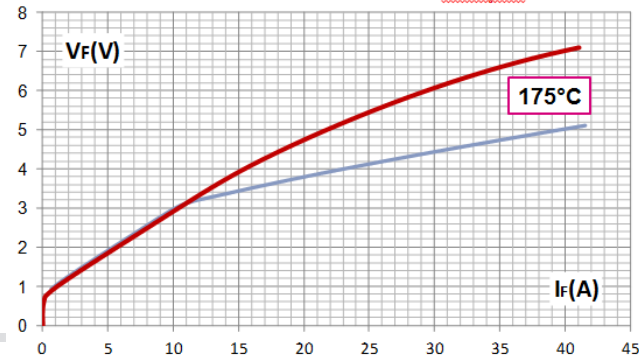
Table 2. Gen 1 diodes (limiting values at 25 °C unless otherwise specified)

Symbol	Gen 1	Parameter	Value	Unit
V_{RRM}		Repetitive peak reverse voltage	600	V
$I_{F(RMS)}$		Forward rms current	18	A
$I_{F(AV)}$		$T_c = 125$ °C, $\delta = 0.5$	6	A
		$t_p = 10$ ms sinusoidal, $T_c = 25$ °C	27	A
I_{FSM}		$t_p = 10$ ms sinusoidal, $T_c = 125$ °C	22	A
		$t_p = 10$ μ s square, $T_c = 25$ °C	110	A
		$\delta = 0.1$, $T_c = 110$ °C, $T_J = 150$ °C	27	A
I_{FRM}		Repetitive peak forward current	27	A
T_{stg}		Storage temperature range	-55 to +175	°C
T_J		Operating junction temperature range	-40 to +175	°C

Better Vf in surge conditions

Competition benchmark
6 A diode

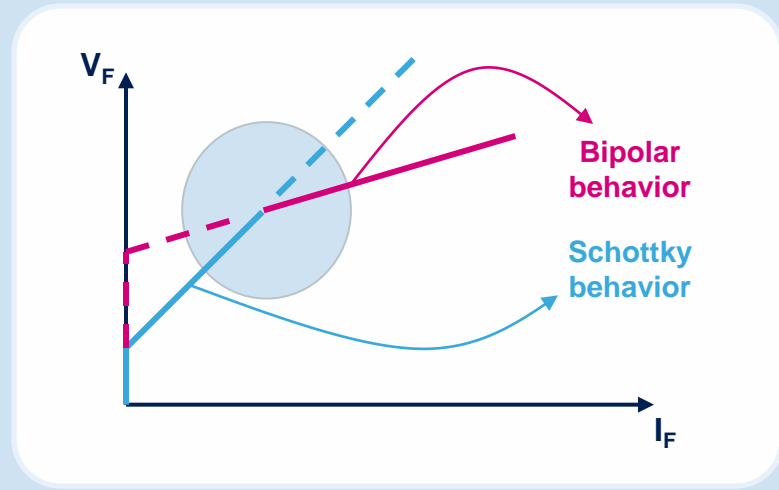
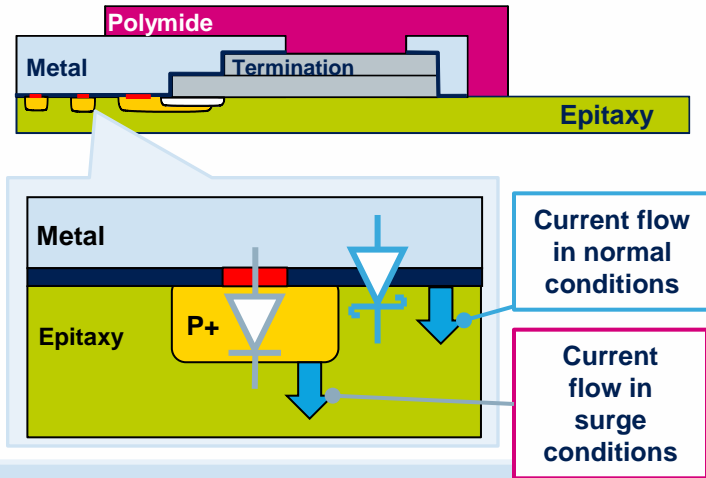
— ST 6A G2
— « compet C » 6A G2



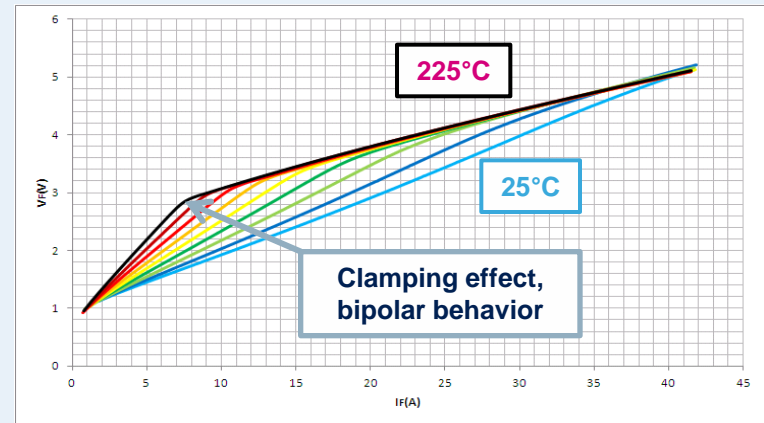
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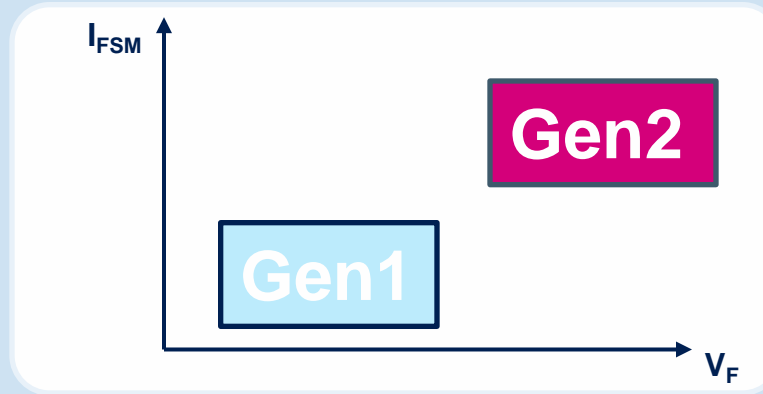
From 600V Generation 1 to 650V Generation 2



- Generation 2 SiC Diode is using a JBS design (Junction-Barrier Schottky)
- The JBS structure overcomes the drawback of positive thermal coefficient present in Generation 1
- The addition of P+ implantation in the Schottky structure creates P/N junctions
- The surge forward current capability can be increased while keeping $T_J < T_{J(MAX)}$



Differences between Generation 1 and Generation 2



600V SiC Diode Generation 1	650V SiC Diode Generation 2
Lower forward conduction losses (linked to V_F)	Higher surge robustness (linked to I_{FSM})
Brings good efficiency levels thanks to the low forward voltage drop	Provides the best trade-off between efficiency and robustness thanks to the high I_{FSM} level
Ideal for use in applications without current surge issues (e.g. solar, UPS)	Ideal for use in applications with high current surge constrains (e.g. PFC circuits in server or telecom SMPS)

Gen 2 does NOT replace Gen 1!!

Generation 1 600 V diodes

Single diodes

STPSC xx 06 yy

SiC Schottky

$I_{F(AV)}$

V_{RRM}

Package

4A, 6A, 8A, 10A, 12A



Dual diodes / common cathode

STPSC xx 06 C yy

Common cathode

2 x 10A, 600V



Generation 2 650 V diodes

Single diodes

STPSC xx H 065 yy

SiC Schottky

$I_{F(AV)}$

Gen 2

V_{RRM}

Package

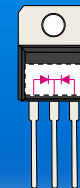
4A, 6A, 8A, 10A, 12A



Dual diodes
Common cathode

STPSCxxH065yy

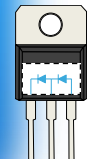
2 x 4A, 650V
2 x 6A, 650V
2 x 8A, 650V
2 x 10A, 650V



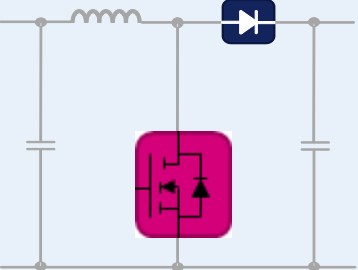
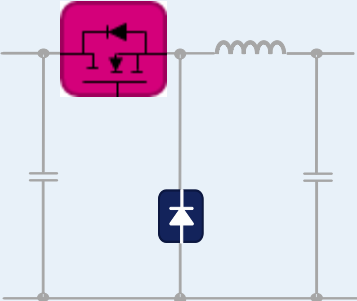
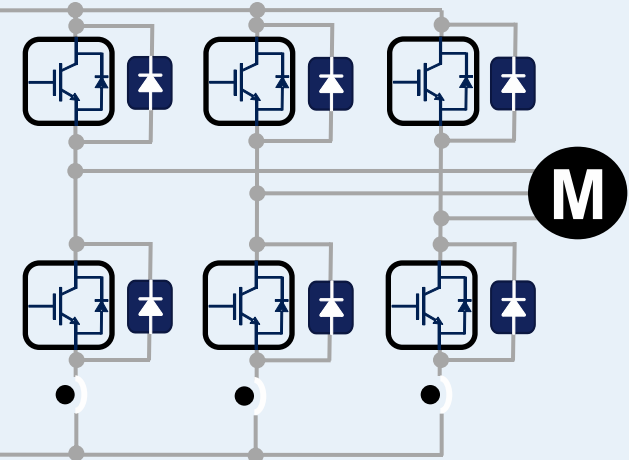
Dual diodes
In series

STPSCxx_H13TI

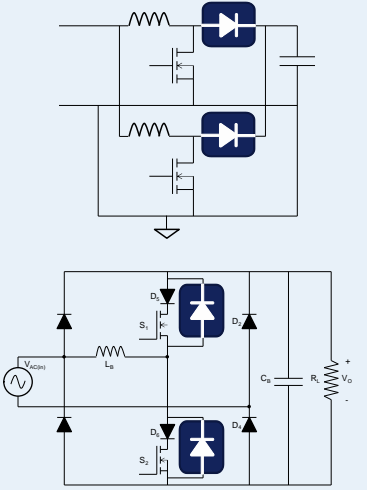
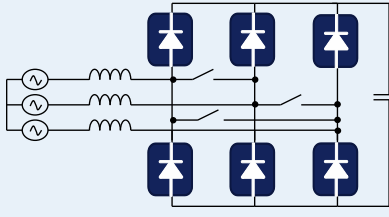
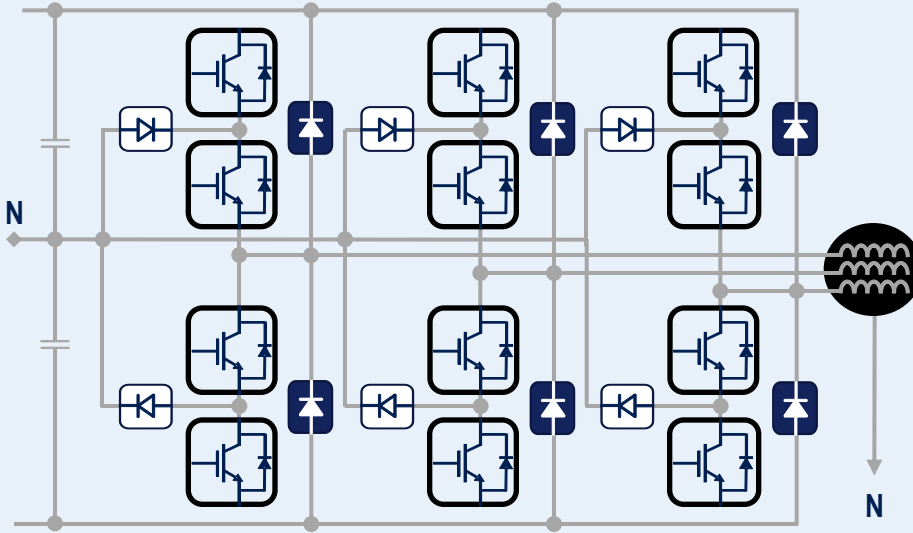


6A, 2 x 650V
8A, 2 x 650V
10A, 2 x 650V



Topologies requiring 1200V SiC MOSFET and 600/650V SiC Diodes

Boost Converter	Buck Converter	3-Phase Motor Drive
		
<p>SCT30N120 1200V SiC MOSFET</p> <p>STPSCxxH065yy 650V Generation 2 SiC Diodes</p> <p>STPSCxxH12yy 1200V Generation 2 SiC Diodes</p> <p>In continuous current conduction mode, low recovery charge = higher efficiency</p>	<p>SCT30N120 1200V SiC MOSFET</p> <p>STPSCxxH065yy 650V Generation 2 SiC Diodes</p> <p>STPSCxxH12yy 1200V Generation 2 SiC Diodes</p> <p>In continuous current conduction mode, low recovery charge = higher efficiency</p>	<p>STPSCxxH065yy 650V Generation 2 SiC Diodes</p> <p>Generation 2 SiC Diodes Additional free-wheeling diodes to lower free-wheeling power losses</p>

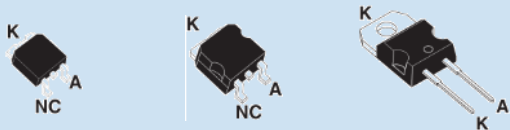
Topologies requiring 600V/650V/1200V SiC diodes

Interleaved or Bridgeless PFC	3-Phase PFC	3-Phase Photo Voltaic Inverter
		
<p>SCT30N120 1200V SiC MOSFET</p> <p>STPSCxxH065yy 650V Generation 2 SiC Diodes</p> <p>STPSCxxH12yy 1200V Generation 2 SiC Diodes</p>	<p>SCT30N120 1200V SiC MOSFET</p> <p>STPSCxxH065yy 650V Generation 2 SiC Diodes</p> <p>STPSCxxH12yy 1200V Generation 2 SiC Diodes</p>	<p> STPSCxx06yy 600V Generation 1 SiC Diodes</p> <p> STPSCxxTH13yy 650V Generation 2 SiC Diodes in Series</p> <p>SiC free-wheeling diodes to lower free-wheeling power losses</p>

Generation 1 - 600V Diodes

Single Diodes

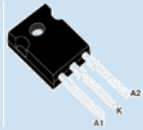
STPSC xx 06 yy



STPCS	ST SiC Schottky
xx	$I_{F(AV)}$ 4A, 6A, 8A, 10A, 12A
06	V_{RRM} 600V
yy	TO-220AC, DPAK, D ² PAK

Dual Diodes / Common Cathode

STPSC xx 06 C yy



C	Common Cathode
xx	$I_{F(AV)}$ 2 x 10A
yy	TO-247

Generation 2 - 650V Diodes

Single Diodes

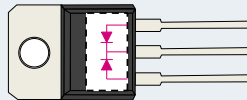
STPSC xx H 065 yy



STPCS	ST SiC Schottky Gen2 (H)
xx	$I_{F(AV)}$ 4A, 6A, 8A, 10A, 12A
065	V_{RRM} 650V
yy	TO-220AC, DPAK, D ² PAK

Dual Diodes / Common Cathode

STPSC xx H 065 C yy

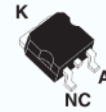


C	Common Cathode
xx	$I_{F(AV)}$ 2 x 4A, 6A, 8A, 10A
yy	TO-220AB, TO-247

Generation 2 - 1200V Diodes

Single Diodes

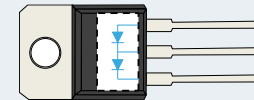
STPSC xx H 12 yy



STPCS	ST SiC Schottky Gen2 (H)
xx	$I_{F(AV)}$ 6A
12	V_{RRM} 1200V
yy	DPAK HV 2 Leads

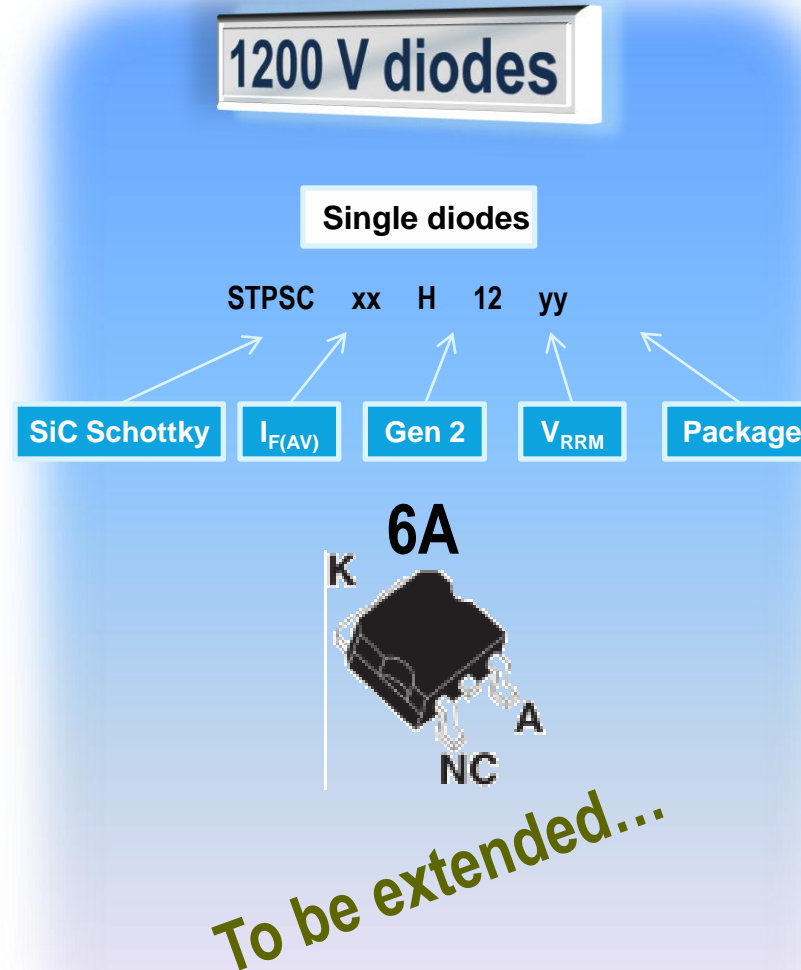
Dual Diodes in Series

STPSC xx TH 13 TI yy



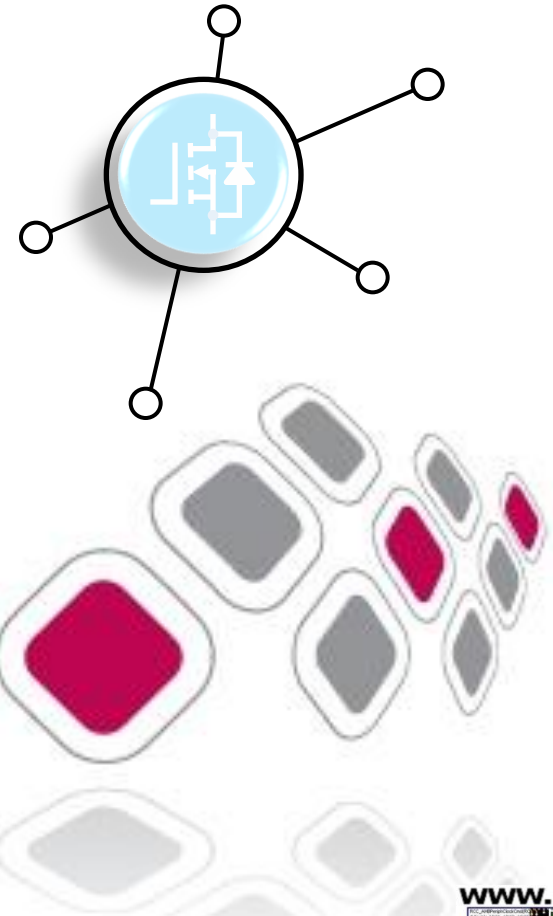
TI	Diodes in Series
xx	$I_{F(AV)}$ 2 x 6A, 8A, 10A
yy	TO-220AB Insulated

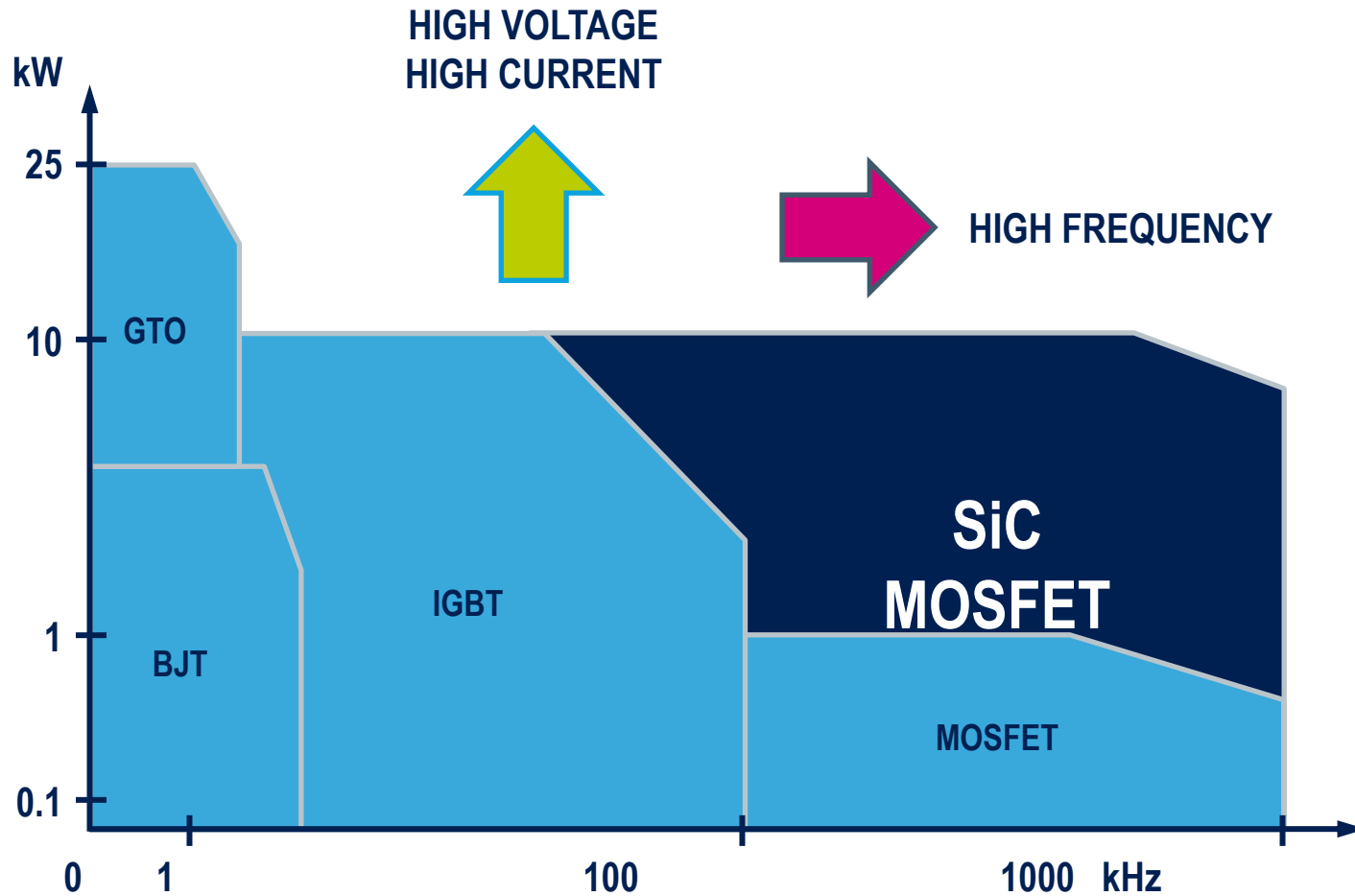
Remember: Generation 2 **does not** replace Generation 1

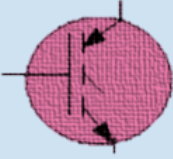
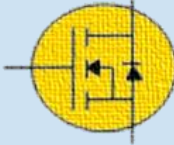



SiC MOSFET

Positioning





Benchmark Summary			
Power Transistor / Parameter	IGBT	MOSFET	SiC MOSFET
Symbol			
Control parameter	Voltage	Voltage	Voltage
Control power	Low	Low	Low
Control circuit	Simple	Simple	Simple
On-resistance	Vce_sat ($\approx 2V$)	Medium	Low
Switching speed	Medium	Fast	Fast
Switching loss	Medium	Low	Extremely low
Load current	High	Medium	High
Operating junction temperature	Up to 175°C	Up to 150°C	> 200°C

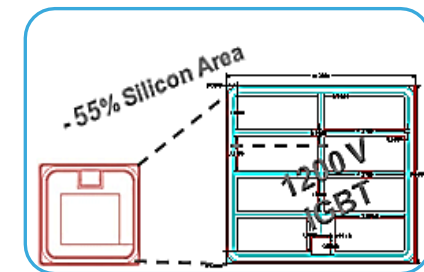
Switching power losses: SiC MOSFET vs. IGBT

SiC MOSFET vs. best in class IGBT

Parameters and Conditions	Die size	V_{on} typ. (V) @ 20A, 25°C	V_{on} typ. (V) @ 20A, 175°C	E_{on} (μJ) @ 20A, 900V 25°C / 175°C	E_{off} (μJ) @ 20A, 900V 25°C / 175°C	E_{off} 25°C / 175°C difference (%)
SiC MOSFET	0.45	2	2.4	725 / 965*	245 / 307	+25% from 25°C to 175°C
IGBT	1.00	1.95	2.35	2140 / 3100	980 / 1850	+90% from 25°C to 175°C

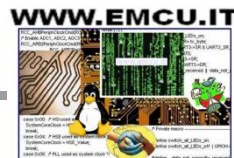
SiC MOSFET

- Results have been measured on SiC MOSFET engineering samples
- SiC device: SCT30N120, 1200V, 45A, 80mΩ, N-channel MOSFET
- IGBT device: best in class field-stop IGBT
- SiC switching power losses are considerably lower than the IGBT ones
- At high temperature, the gap between SiC and IGBT is insurmountable
- SiC MOSFET is the optimal fit for high-power, high-temperature applications



SiC die size compared to IGBT

* E_{off} measured using the SiC intrinsic body diode



SCT30N120

Silicon carbide Power MOSFET
45 A, 1200 V, 80 mΩ, N-channel in HiP247™ package

Datasheet - preliminary data

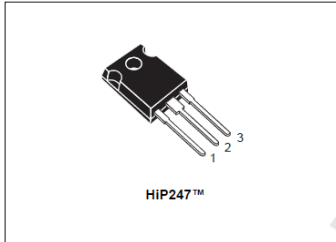
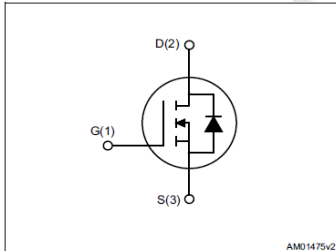


Figure 1. Internal schematic diagram



Features

- Very tight variation of on resistance vs. temperature
- Slight variation of switching losses vs. temperature
- Very high temperature operation capability (200 °C)
- Very fast and robust intrinsic body diode
- Low capacitance
- Easy to drive

Applications

- Solar inverters, UPS
- Motor drives
- High voltage DC-DC converters
- Switch mode power supply

Description

This silicon carbide Power MOSFET is fabricated exploiting the advanced and innovative properties of wide bandgap materials. These include unsurpassed on-resistance per unit area and very good switching performance almost independent of temperature. The outstanding thermal properties of the SiC material together with the adoption of the proprietary HiP247™ package allows designers to use an industry standard outline with significantly improved thermal capability. All these features make the device perfectly suitable for high-efficiency and high power density applications.

Table 1. Device summary

Order code	Marking	Package	Packaging
SCT30N120	SCT30N120	HiP247™	Tube

Note: The device meets ECOPACK standards, an environmentally-friendly grade of products commonly referred to as "halogen-free". See Section 3: Package mechanical data.

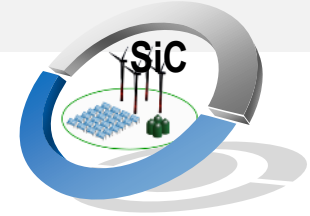
24 April 2013

DocID023109 Rev 2

1/9

This is preliminary information on a new product now in development or undergoing evaluation. Details are subject to change without notice.

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1200V SiC MOSFET

$$V_{BR} > 1,200 \text{ V}$$

$$I_n = 45 \text{ A}$$

$$R_{on(typ)} < 80 \text{ m}\Omega$$

$$Q_g(typ) < 105 \text{ nC}$$

Gate Driving Voltage = 20 V

HiP247 Package : $T_{jmax} = 200 \text{ }^\circ\text{C}$



Switching power losses vs IGBT

SiC MOSFET vs. best in class IGBT

Device	Von typ (V) (@ 25° C, 20A)	Von typ (V) (@ 175° C, 20A)	Eon (uJ) @ 20A, 900V 25° C/175° C	Eoff(uJ) @ 20A, 900V 25° C/175° C	Chip size
SCT30N120 (ST SiC MOSFET)	2	2.4	725/ 965(*)	245/307	0.45
IGBT	1.95	2.35	2140/3100	980/1850	1

+ 30 % at 175°C

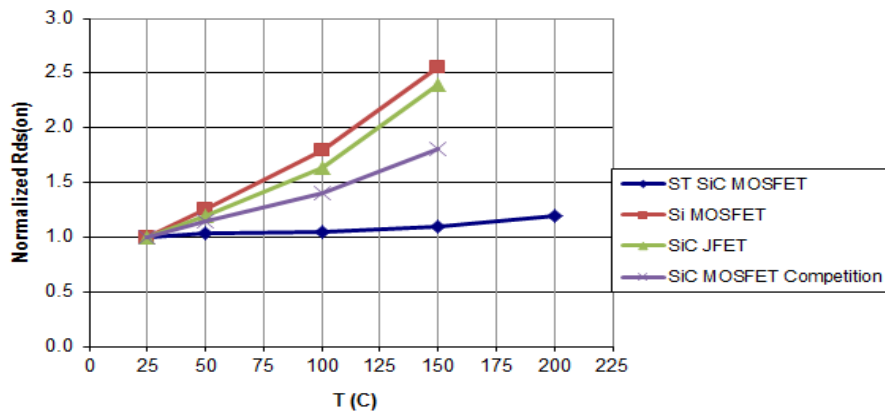
+ 90 % at 175°C

Outstanding RDson/mm² with SiC-Mosfets

Part number	BV	Package	Ron (mΩ)	Ron*A (mΩ*cm²)
SCT30N120 (ST SiC MOSFET)	1200V	HiP247	80	11.5!
STWN120K5	1200V	TO 247	320	220
IGBT 25A (Vcesat=2.05V@25°C)	1200V	TO 247	80 (*)	20

At high temperatures “simply the best”

Normalized ON resistance vs Temperature



Sorry, 99.3% Efficiency?

- Complete solution of boost inverter 4kW from 400-600Vdc to 800Vdc

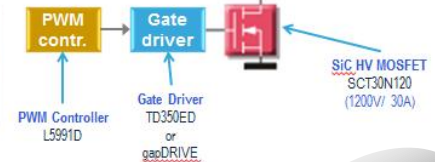
new

- Tested also with GapDrive with same results
- Very appreciated demo – used on several events already
 - PCIM, Daimler day, Paris techday, FAE training
- Final PCB revision arrived
- SiC 1200V diode now mature!



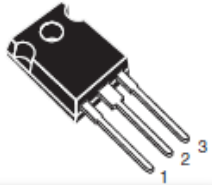
Silicon carbide diodes
2x STPSC6H12B
(1200V/ 6A)

Input voltage (VDC)	Pout (W)	Eff total (%)	Eff without AUX (%)	Heatsink temperature
600	2094	99.11	99.29	57.5C

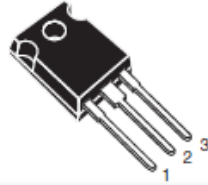


Legend:

- In production
- Coming soon
- Coming next



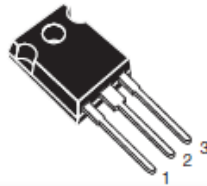
SCT30N120
1200V / 80mΩ
April 2014



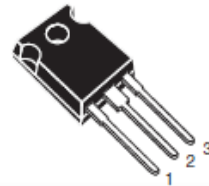
1700V / 300mΩ
Sept. 2014



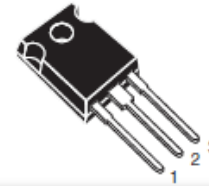
2nd Generation
1700V / 40mΩ
Q1 2015



SCT20N120
1200V / 200mΩ
April 2014



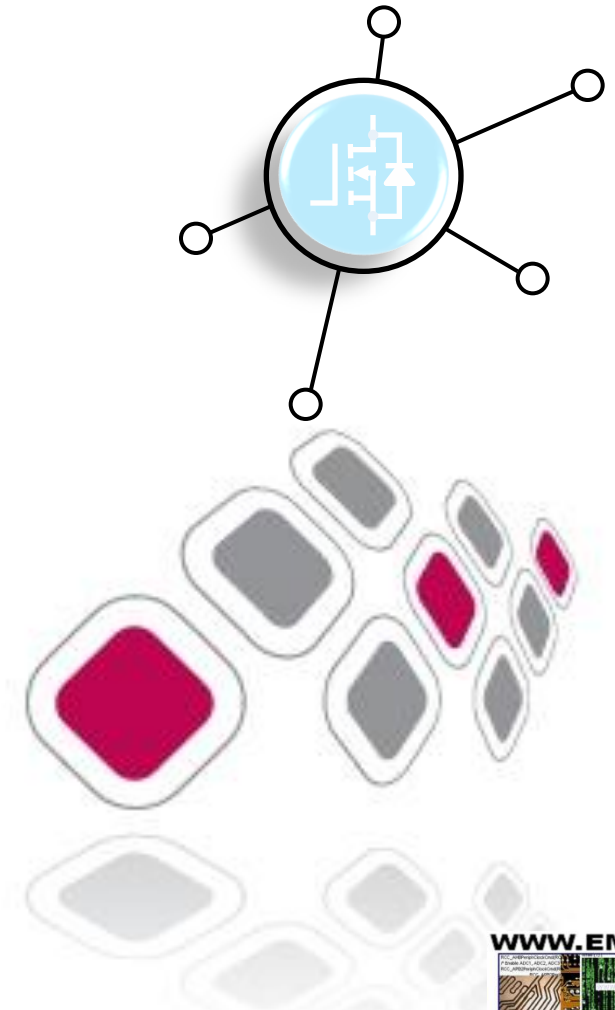
SCT20N170
1700V / 150mΩ
June 2014



2nd Generation
1200V / 50mΩ
Q4 2014

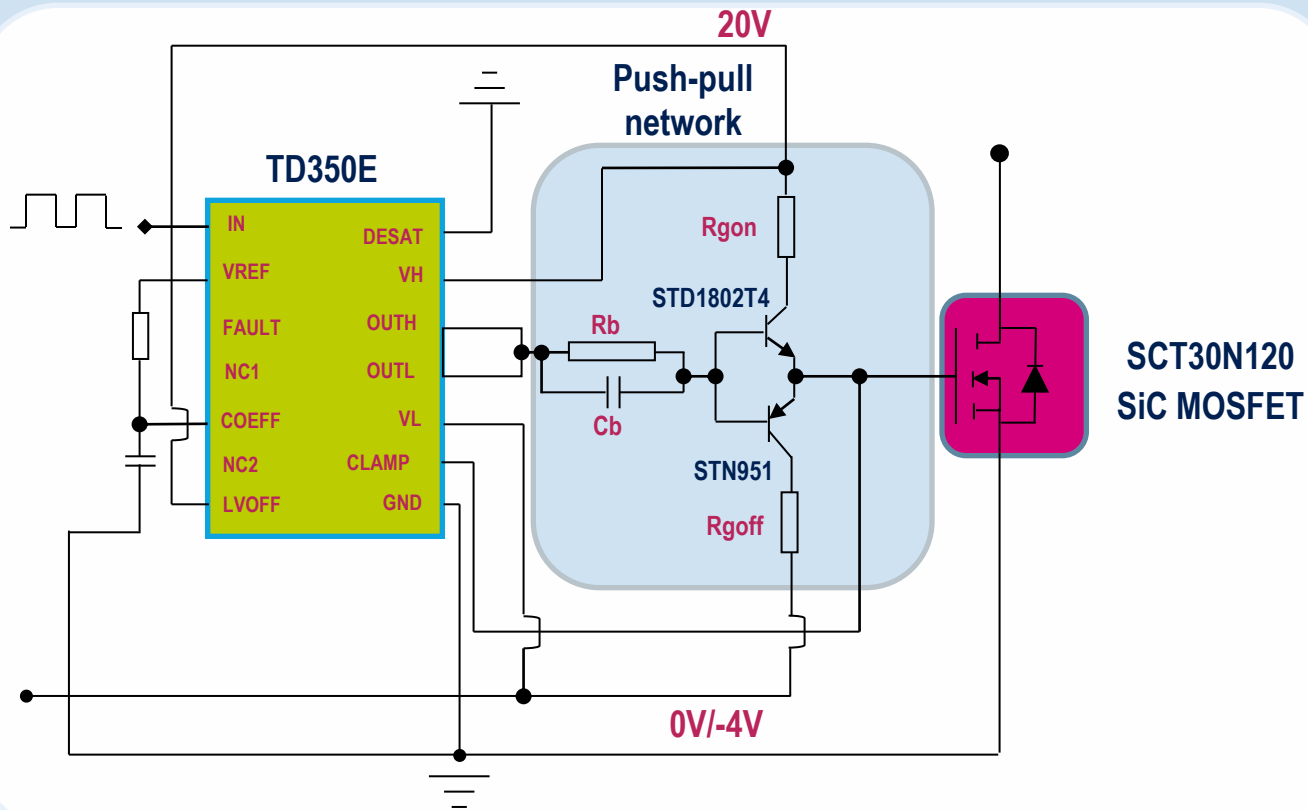


How to drive a SiC MOSFET



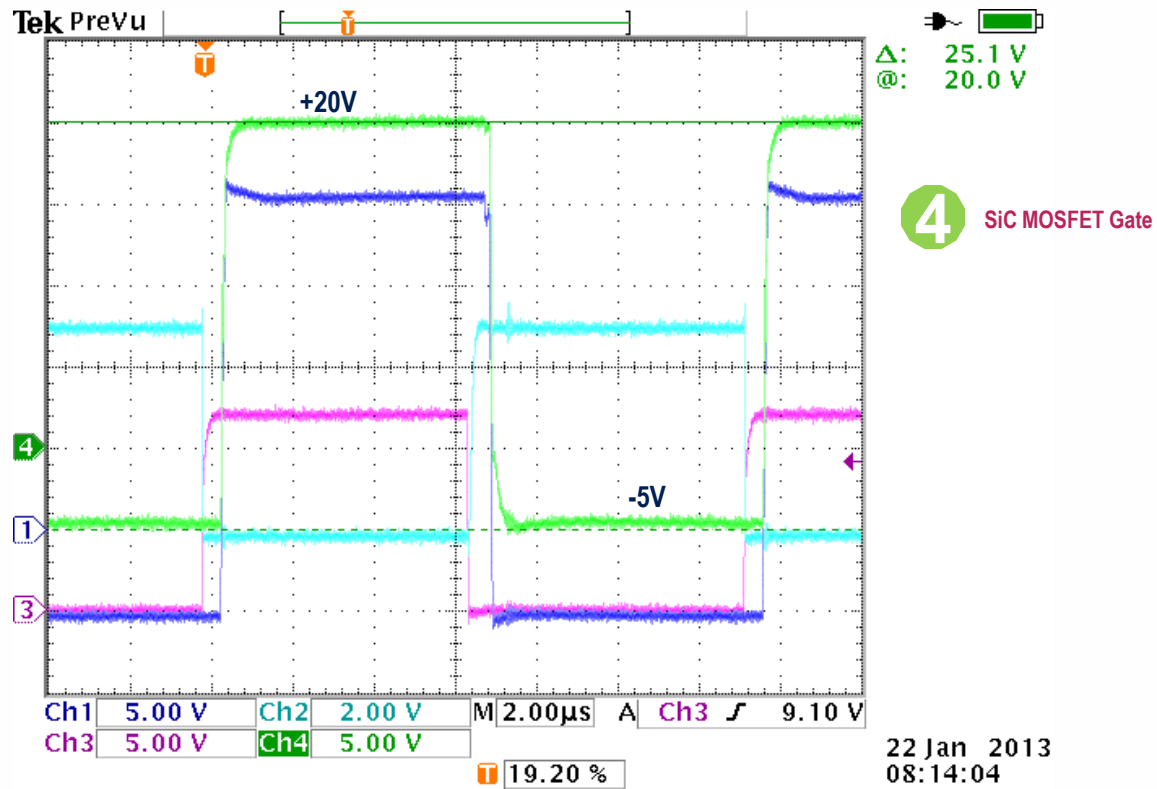
Driving SiC MOSFET with TD350E

The ST **TD350E** is an advanced gate driver for IGBTs and power MOSFETs. To drive a SiC MOSFET, simply add an external push-pull network to increase current capability.



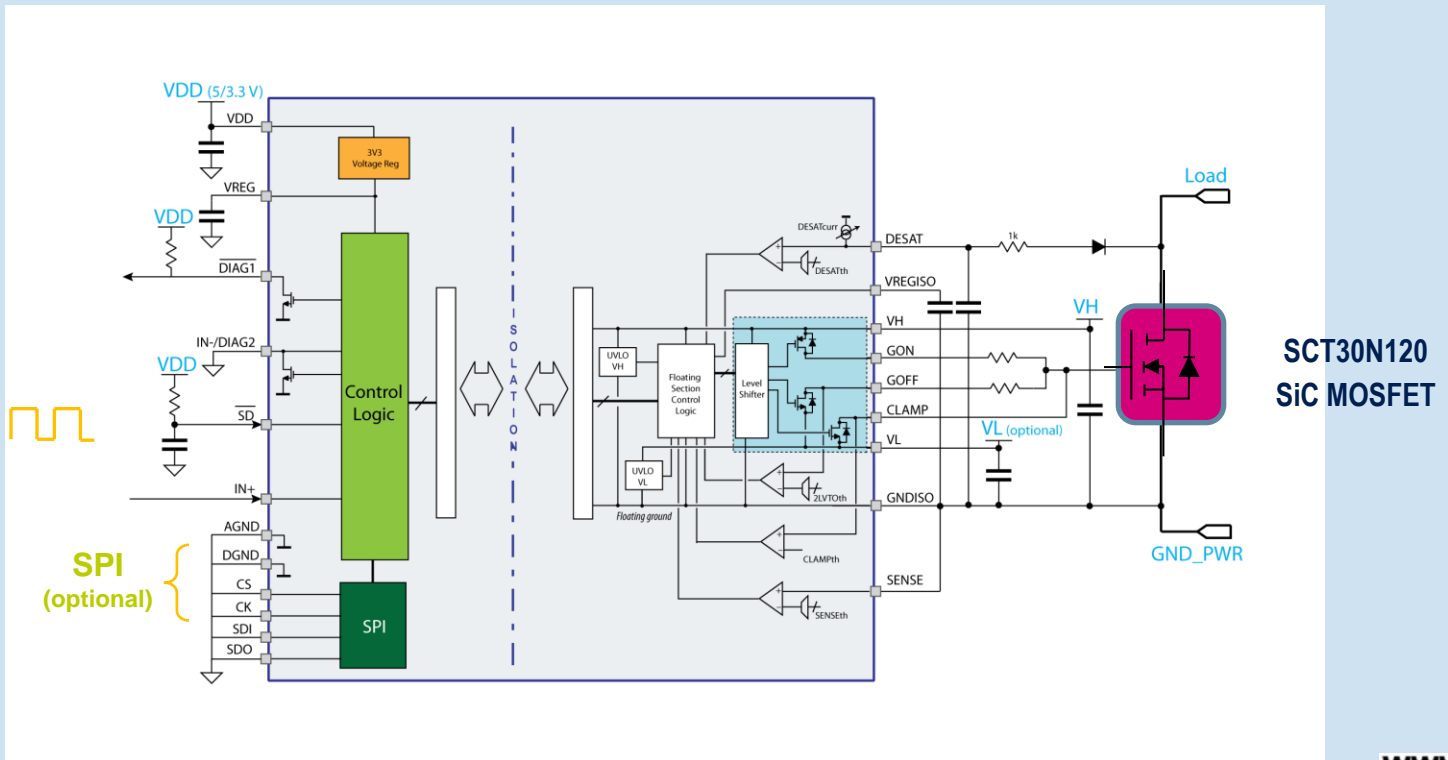
Driving SiC MOSFET

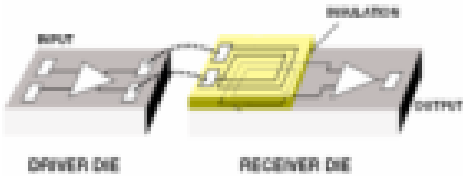
For its driving, SiC MOSFET requires a +20V/-5V gate signal.



Driving SiC MOSFET with GapDrive

The **GapDrive** is a 4000V galvanic isolated gate driver from ST, for IGBTs and power MOSFETs.





Market position

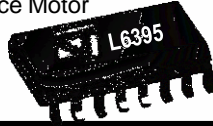
- Today Market Share ~25%
- Leadership in Appliances (~80% MS in EMEA)

HA Leader 25% MS



L6395 smartDRIVE

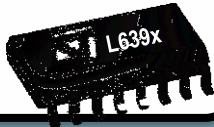
- Optimized for Switched-Reluctance Motor



More than 50Mpcs sold in 2011

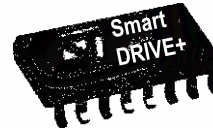
smartDRIVE L639x, L638xE

- Easy sensor-less Driver optimized for FOC



smartDRIVE+

4 A Driver 10x power & Speed



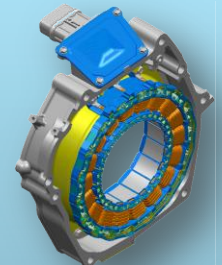
gapDRIVE

- Galvanic Isolated Gate Driver with SPI & Advanced Protection

gapDrive/Lite

- Lite version with selected features
- Smaller size
- Lower cost

Industrial Drives & EV
10kW - 200kW



2012

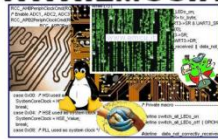
2013

2014

Full production

Design

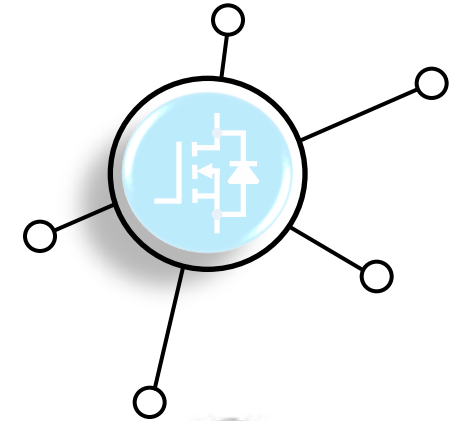
Concept

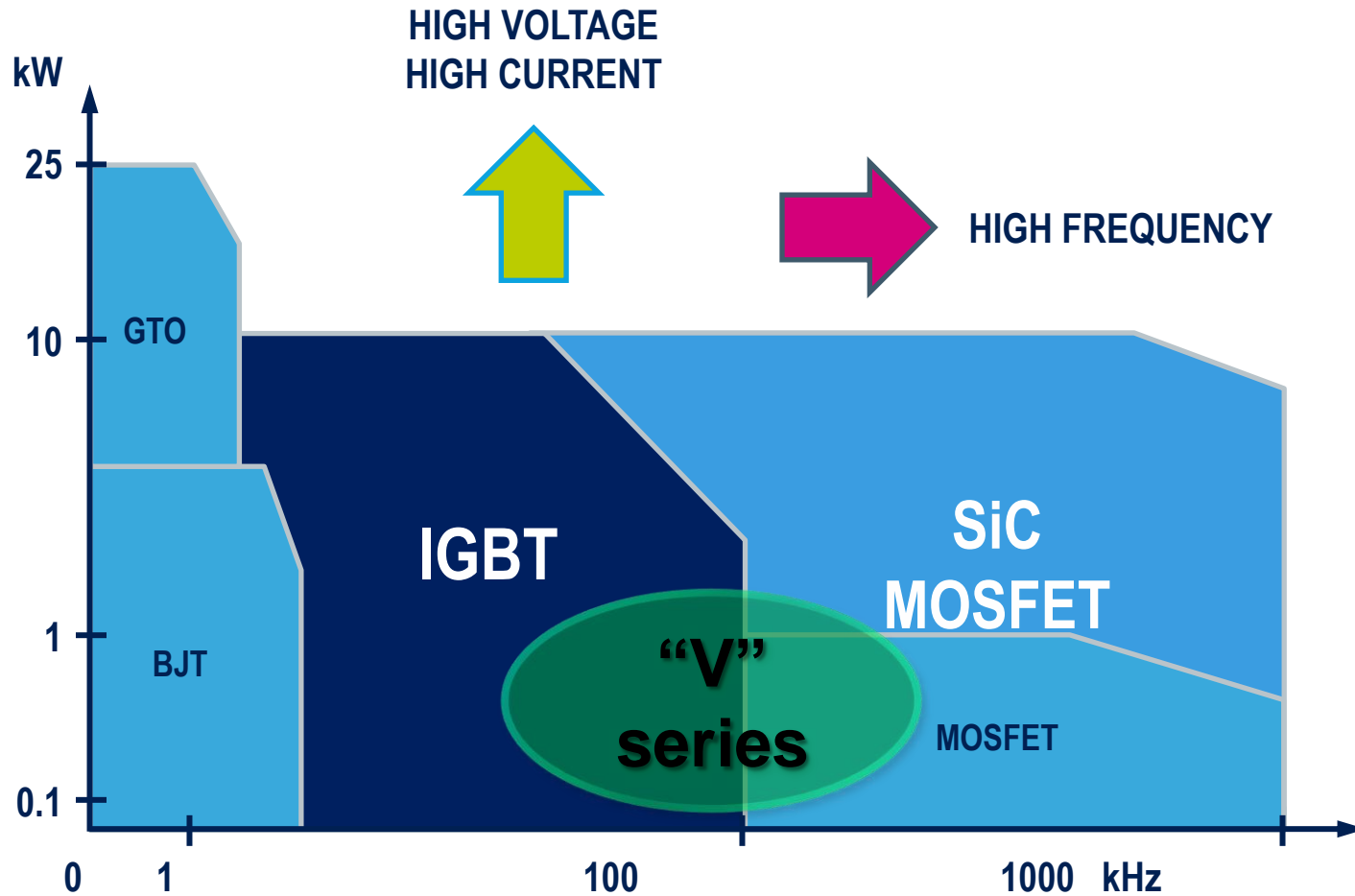


SiC Vs IGBT



600V IGBTs – V & HB Series
New Trench Gate Field Stop
Technology – Very High Speed
Series





New ST Trench Gate Field Stop H, HB, V series

The ideal companion for frequency converter

H

Up to 25 kHz

- Low Frequency Converter
- Motor Control, PFC

HB

Up to 35 kHz

- Medium Frequency Converter, Soft Switching
- PV Inverter; Welding Induction Heating, PFC

V

Above 35 kHz

- High Frequency Converter
- Welding, UPS, PV, PFC



Energy saving



Power scalability

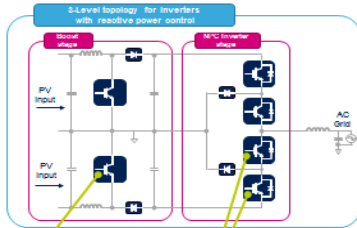


Robustness and reliability

From Silicon to application more than a simply IGBT supplier



Solar inverters



IGBT for boost stage

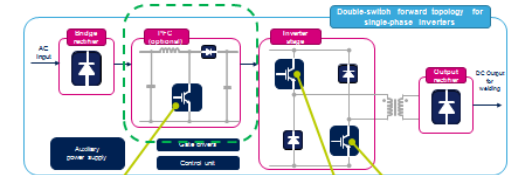
- 600 V trench gate field stop
- STG30V60F
- STG40V60F, STG40H65DFB
- STG60V60F, STG60H65DFB
- STG80V60F, STG80H65DFB

IGBT for inverter stage

- 600 V trench gate field stop
- STG30V60DF
- STG40V60DF, STG40H65DFB
- STG60V60DF, STG60H65DFB
- STG80V60DF, STG80H65DFB



Welding: double-switch forward topology



IGBT for PFC stage

- 600 V trench gate field stop
- STG20V60F, STG20H65DF
- STG30V60F
- STG40V60F, STG40H65DFB
- STG60V60F, STG60H65DFB

IGBT for inverter stage

- 600 V trench gate field stop
- STG20V60F, STG20H65DF
- STG30V60F
- STG40V60F, STG40H65DFB

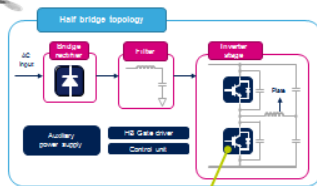


Supported by

IGBT application guidelines



Induction heating

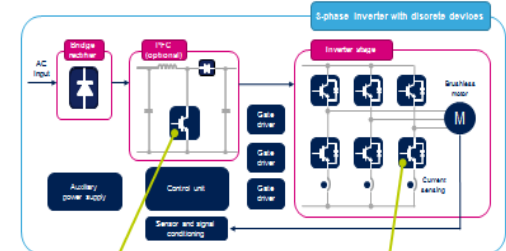


IGBT for inverter stage

- 600 V trench gate field stop
- STG40V60DLF, STG40H65DLFB
- STG60V60DLF, STG60H65DLFB



3-phase Inverter for Brushless Motors



IGBT for PFC stage

- 600 V / 1200 V Trench gate field stop
- STG10M50F
- STG10M60F

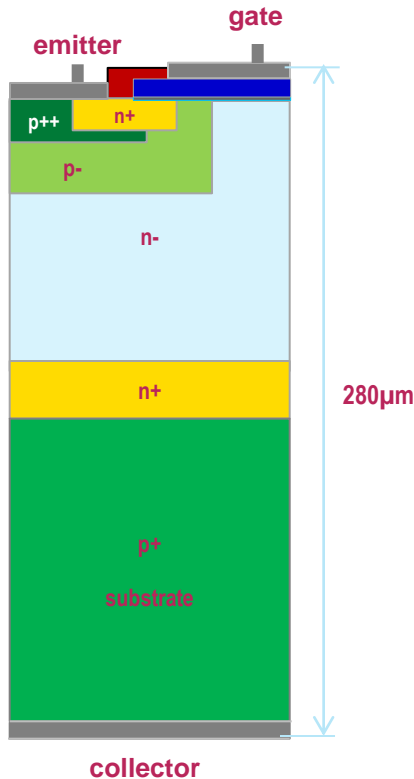
IGBT for inverter stage

- Trench gate field stop
- STG10M60DF

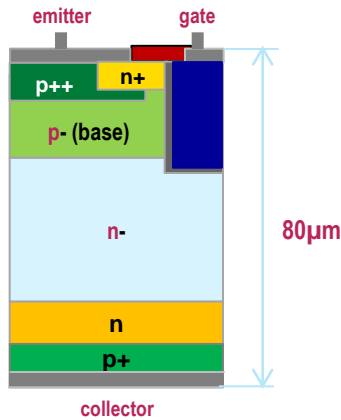


Technology Milestones and Features

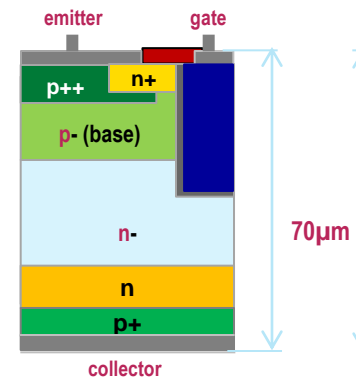
Planar PT



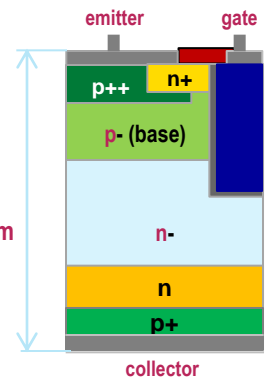
H



V



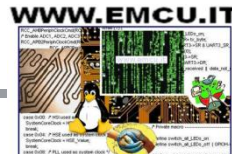
H-B



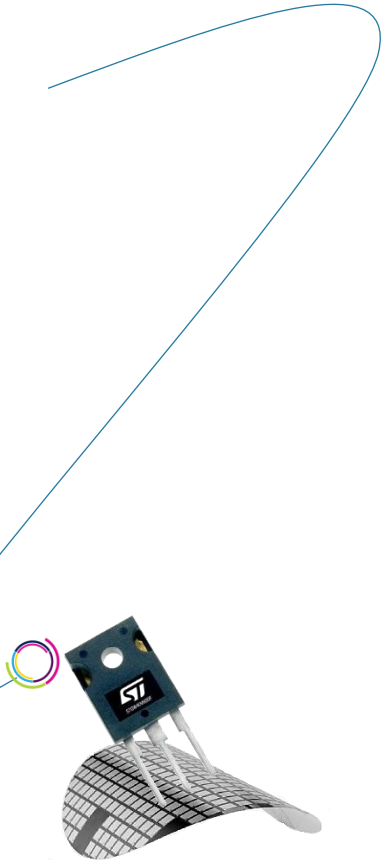
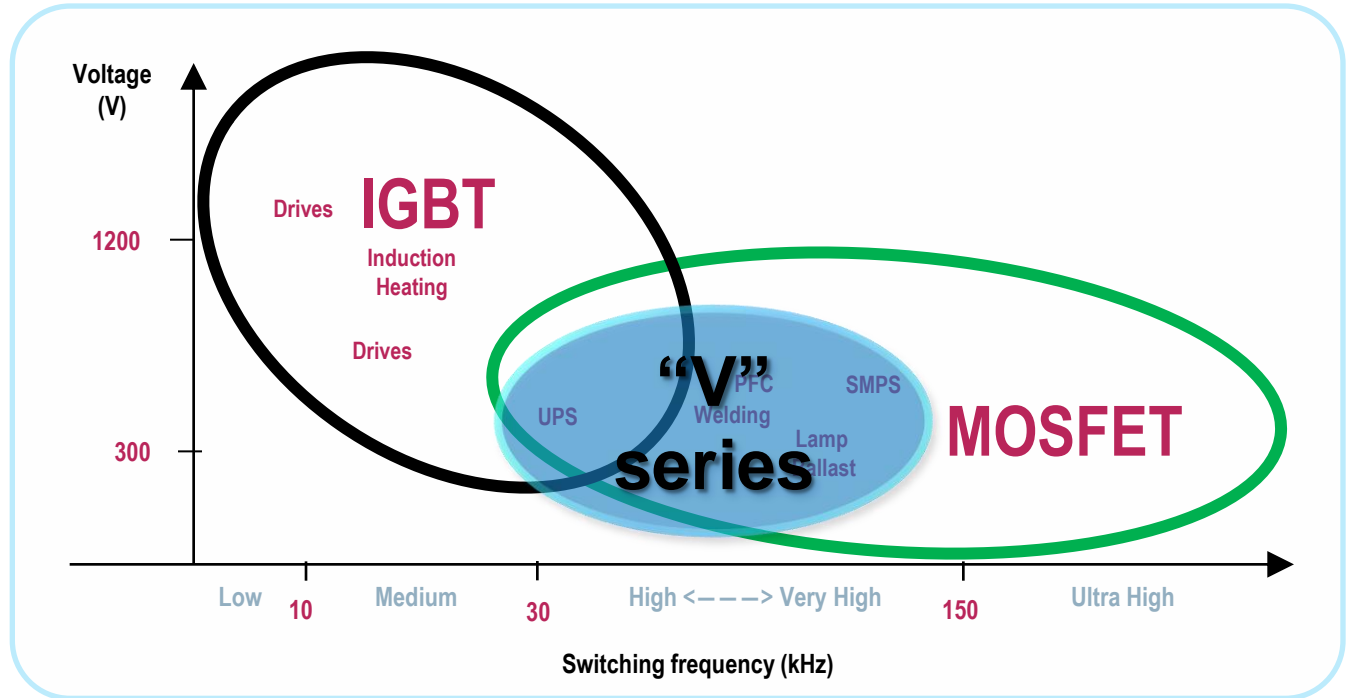
Energy Saving
Extremely low switching-off combined with a low conduction losses.



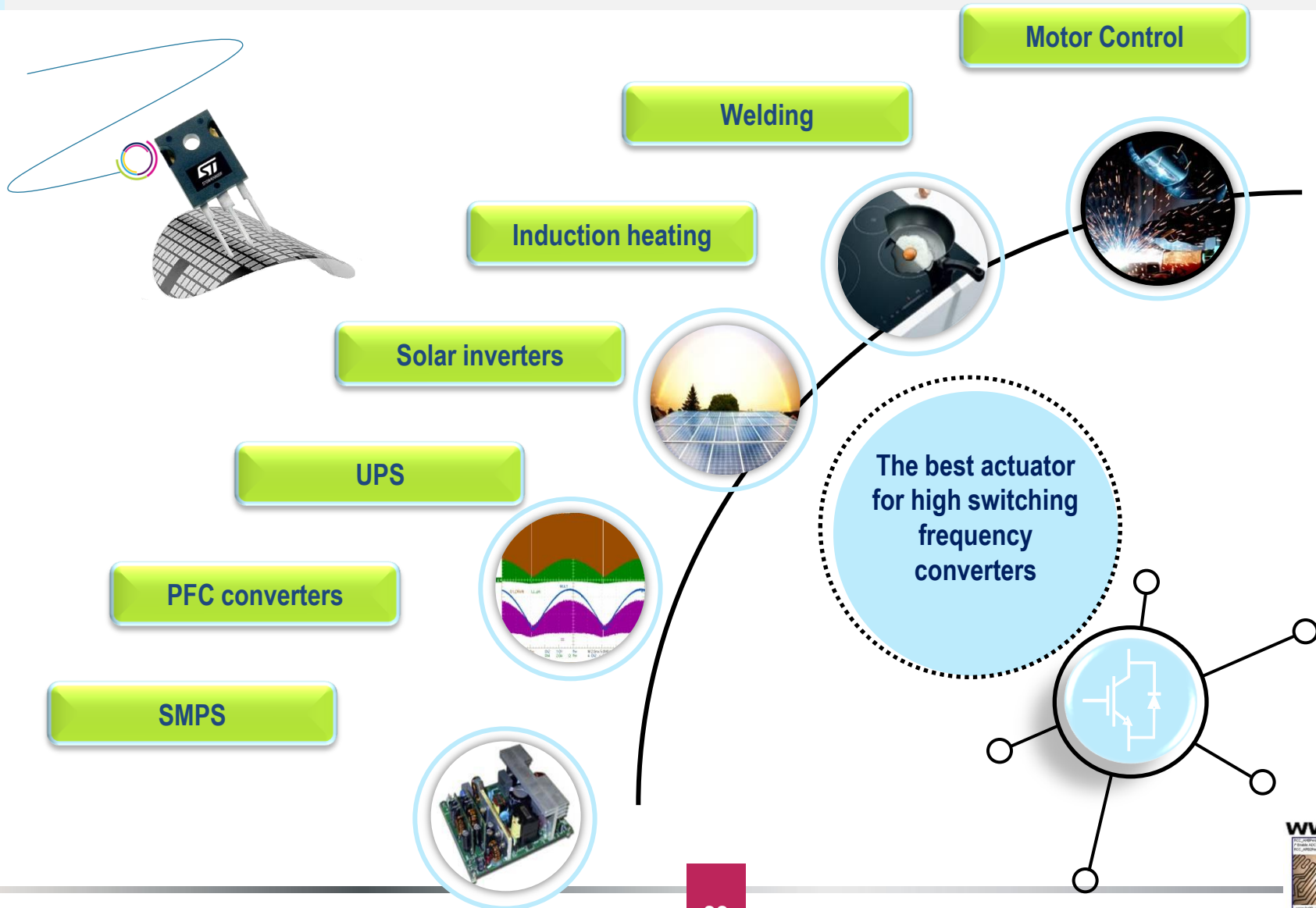
Robustness and Reliability
The Increasing up to 175°C of the max Junction Temperature $T_j(\max)$, Ensures an higher lifetime



Positioning



New IGBT "V" series: developed to bridge the gap between IGBTs and MOSFETs in high frequency hard switching applications above 20kHz



Moving from Planar to Trench Field-Stop

1st time aligned or even better than the market leader!

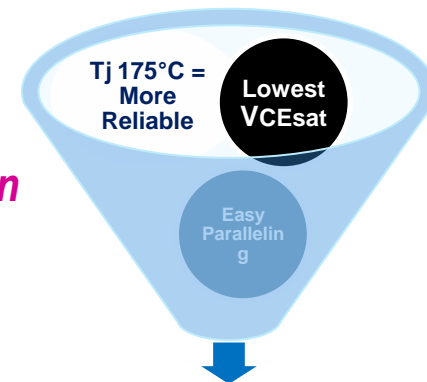


Just a Snapshot:

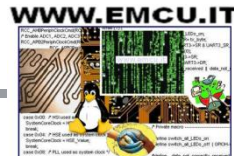
P/N	Package	V_{CES} (V)	I_{CN} (A)	$V_{ce(sat)}$ (V)	E_{off} (uJ)	Diode Option
STGP20V60DF	TO-220	600	20	1.8	130	Very Fast
STGP20V60F	TO-220	600	20	1.8	130	-
STGB30V60DF	D2PAK	600	30	1.85	233	Very Fast
STGB30V60F	D2PAK	600	30	1.85	233	-
STGP40V60F	TO-220	600	40	1.8	411	-
STGW40H65DFB	TO-247	650	40	1.6	450	Very Fast
STGW60V60DLF	TO-247	600	60	1.85	270 ^{a)}	Low Drop
STGW60V60F	TO-247	600	60	1.85	550	-
STGW60H65FB	TO-247	650	60	1.65	650	-
STGWT60H60DLFB	TO-3P	600	60	1.65	130 ^{a)}	Low Drop
STGW80H65DFB	TO-247	650	80	1.65	850	Very Fast
STGY80V60DF	MAX-247	600	80	1.85	850	Very Fast



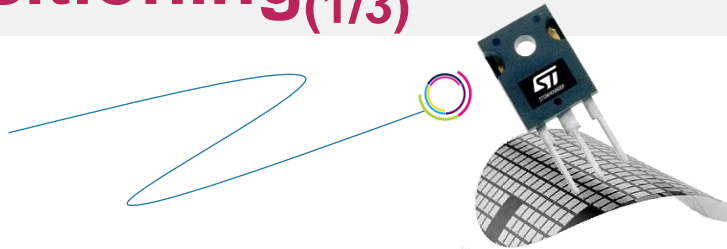
No-Compromise in Efficiency and Competiveness



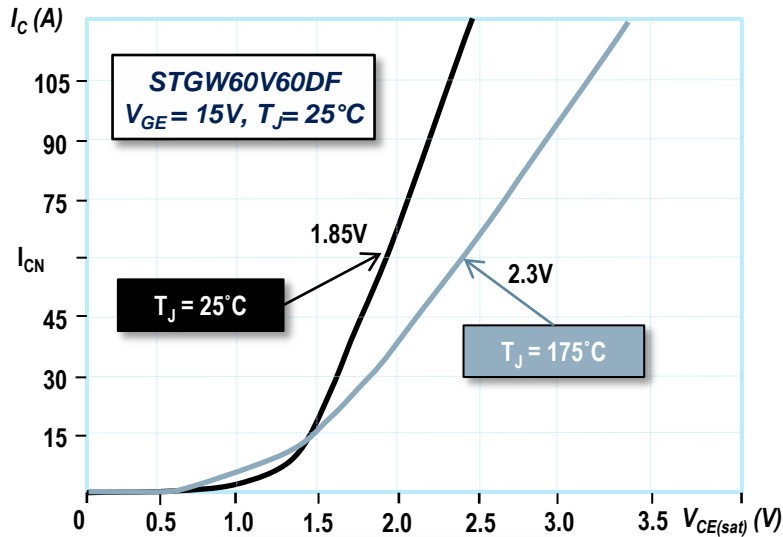
Top Class Efficiency & Reliability



positioning_(1/3)

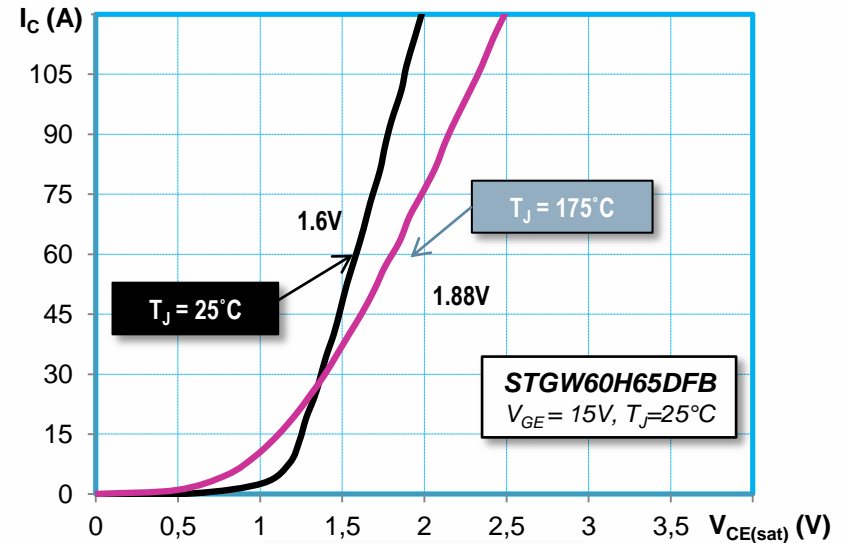


Saturation voltage characteristic



V
Above 35 kHz

Saturation voltage characteristic

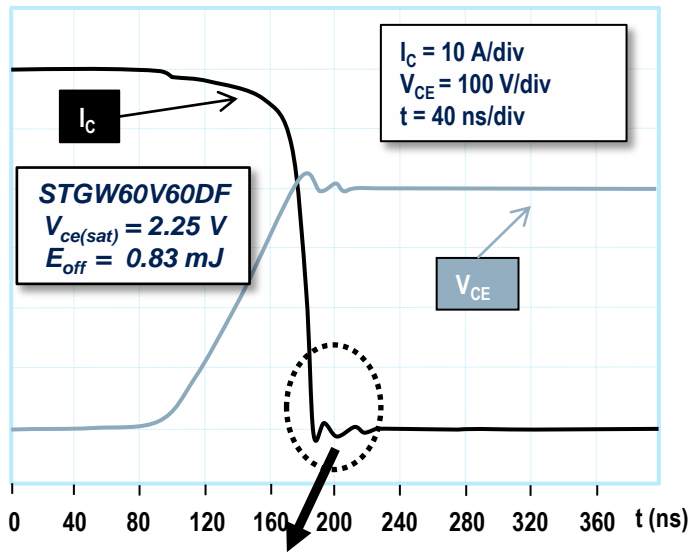


HB
Up to 35 kHz

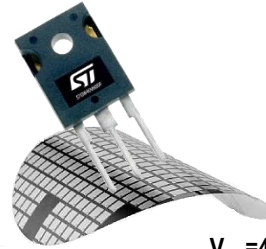
positioning (2/3)

Switching-off waveforms

$V_{CC}=400V, R_G=5\Omega, I_C=I_{CN}, T=150^\circ C$

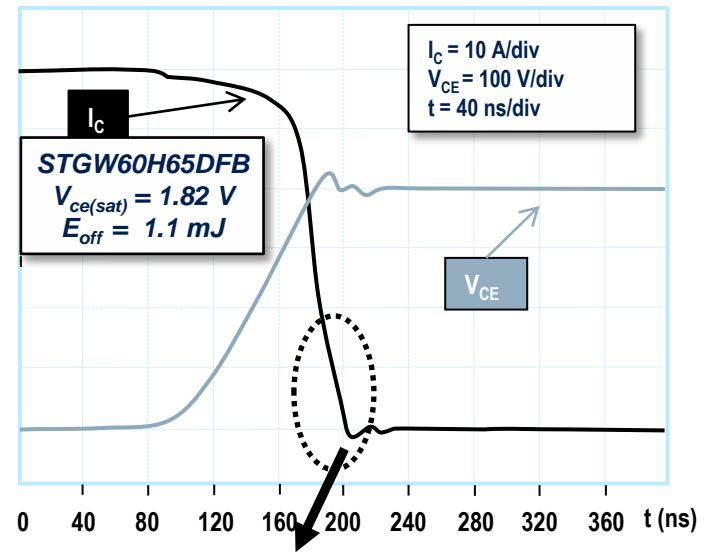


Tail-less switching-off ...
... MOSFET "like" switching-off behavior



Switching-off waveforms

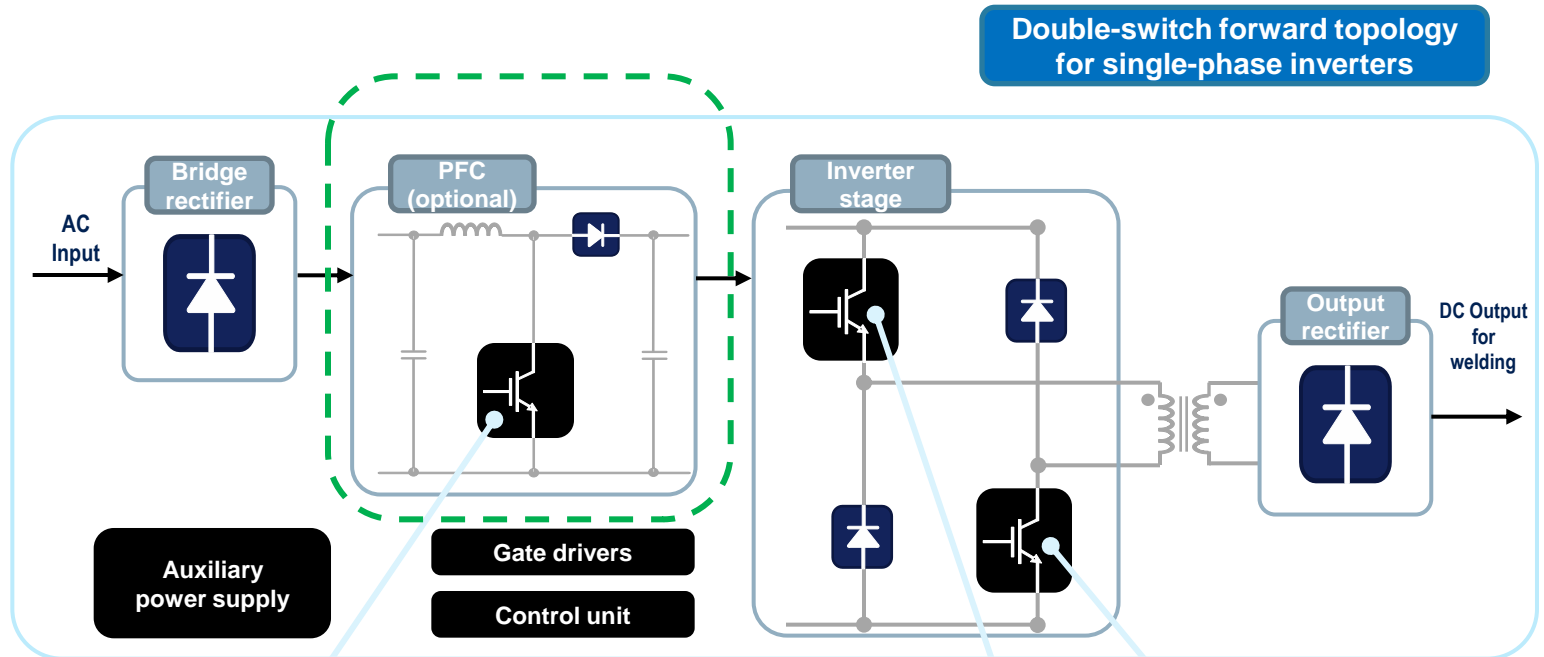
$V_{CC}=400V, R_G=4.7\Omega, I_C=I_{CN}=60A, V_{GE}=15V, T=150^\circ C$



Almost tail-less switching-off ...



Double-switch forward topology



IGBT for PFC stage

- 600 V trench gate field stop
- STGW20V60F, STGW20H60DF
- STGW30V60F
- STGW40V60F, STGW40H65DFB
- STGW60V60F, STGW60H65DFB

IGBT for inverter stage

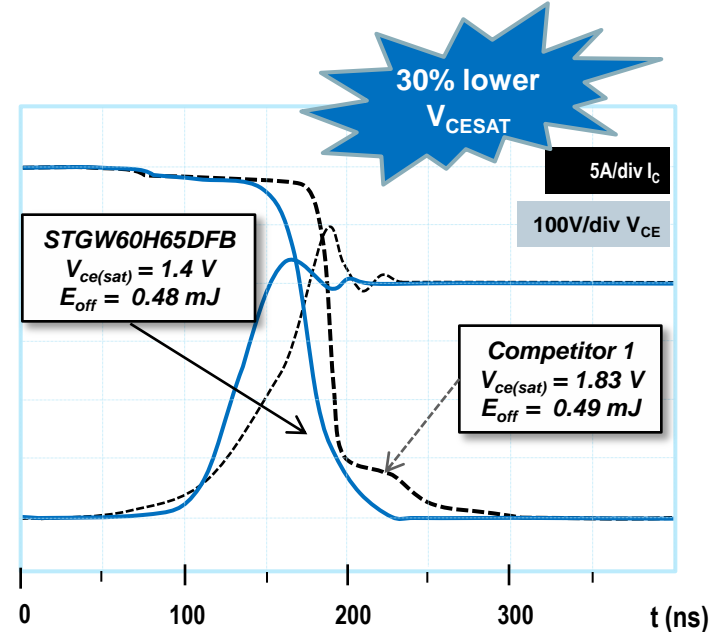
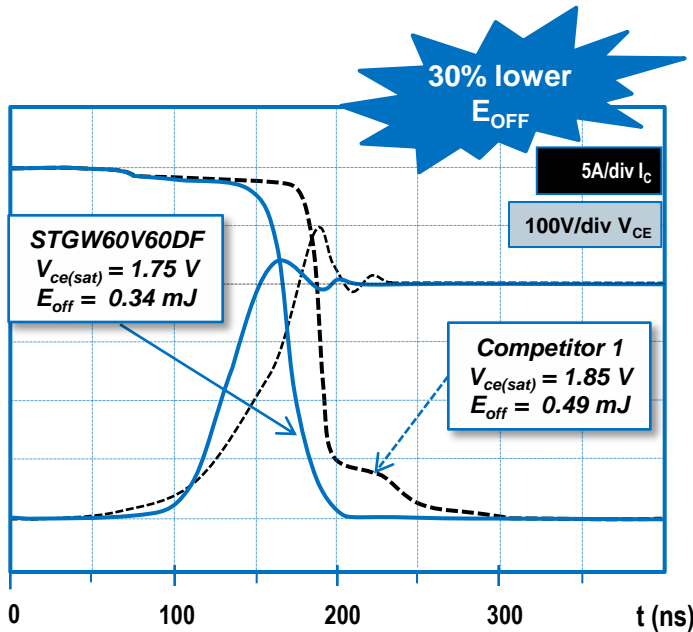
- 600 V trench gate field stop
- STGW20V60F, STGW20H60DF
- STGW30V60F
- STGW40V60F, STGW40H65DFB

Switching-off benchmarking (3/3)

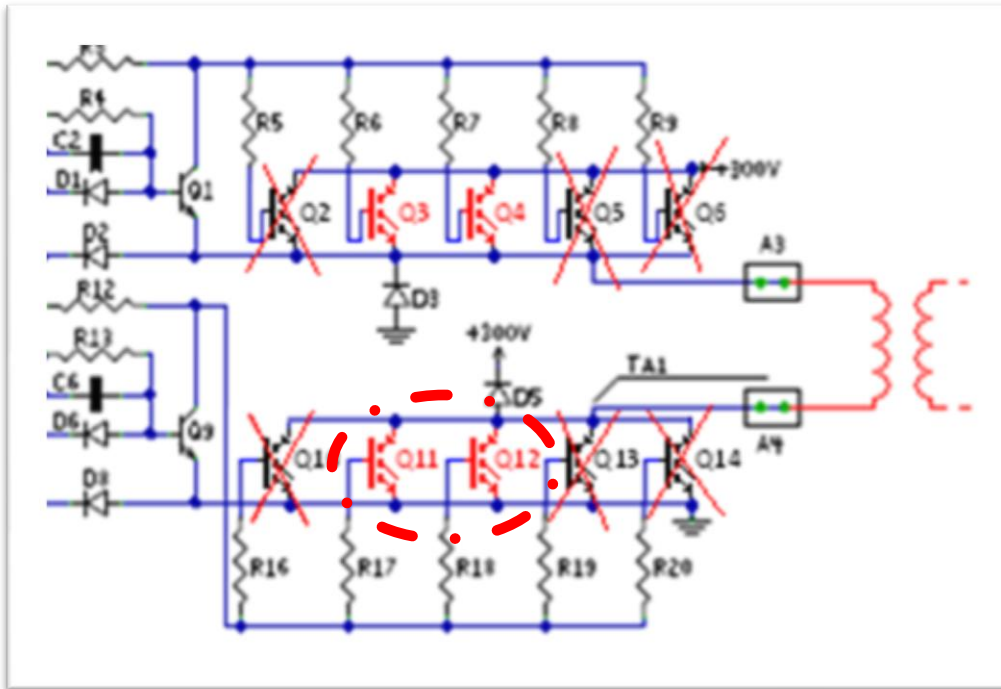


STGW60V60DF vs. Competitor 1

STGW60H65DFB vs. Competitor 1



Test condition: $V_{CC} = 400V$, $R_G = 4.7\Omega$, $I_C = \frac{1}{2} I_{CN} = 30A$, $V_{GE} = 15V$, $T_J = 150^\circ C$



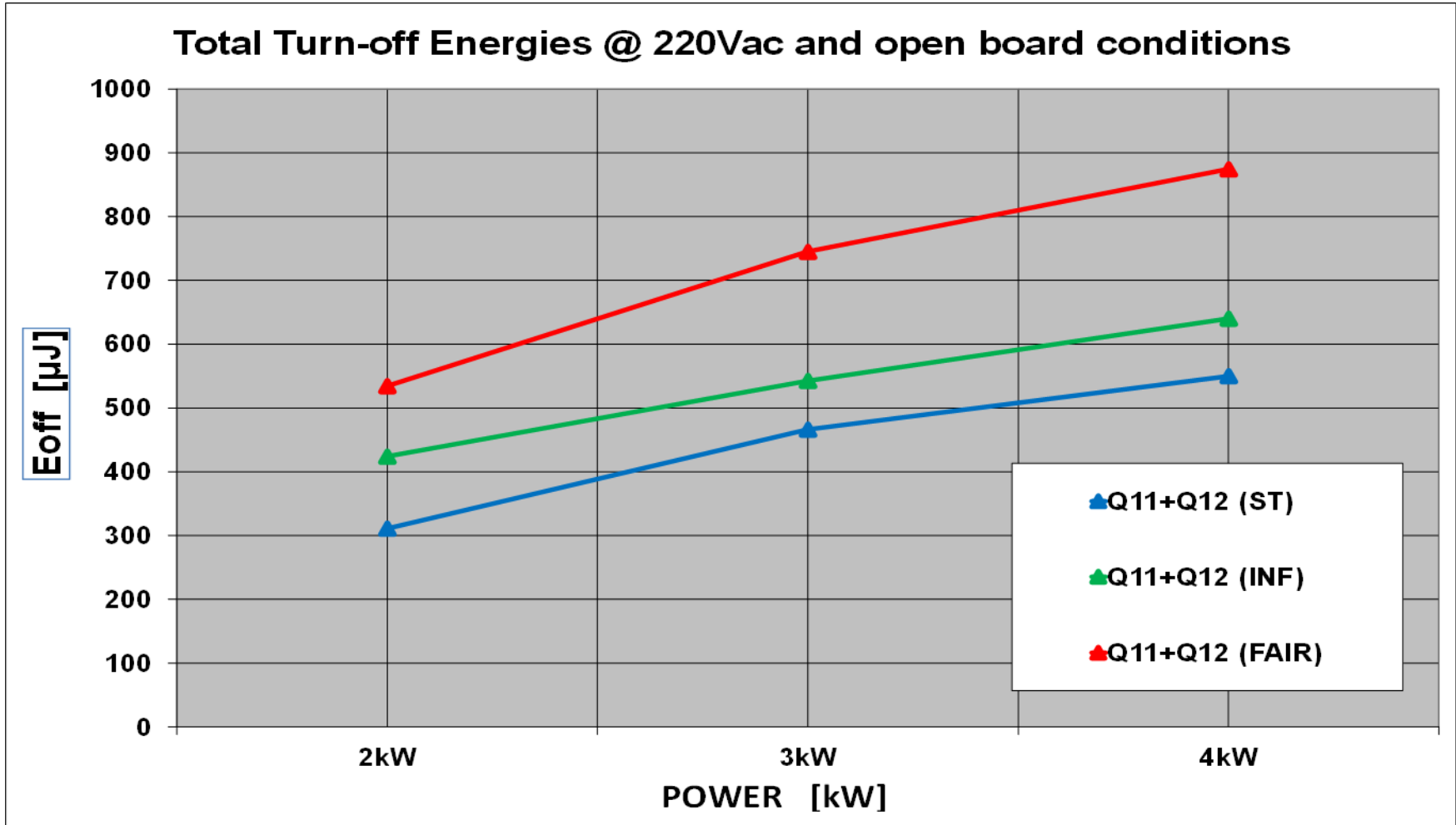
The signal waveforms were acquired and evaluated for two devices Q11 and Q12 which are connected in parallel in the low side part of the double switch forward converter section

Devices under analysis:

STGW40V60DF

Competitor 1

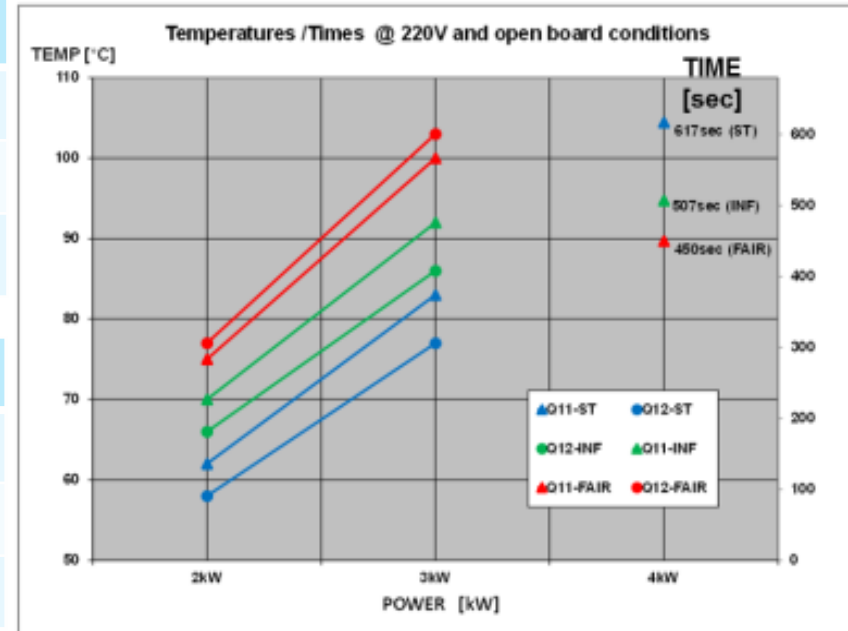
Competitor 2



DEVICE	Pin (W)	Input Ic (A)	PF	T [°C]
STGW40V60DF	2	15.4	0.58	62° - 58°
	3	22.2	0.61	83° - 77°
	~3.8 (MAX)	26.3	0.66	10 min:17 sec

DEVICE	Pin (W)	Input Ic (A)	PF	T [°C]
Competitor 1	2	15.6	0.58	70° - 66°
	3	22.3	0.61	92° - 86°
	~3.8 (MAX)	26.4	0.66	8min:27sec

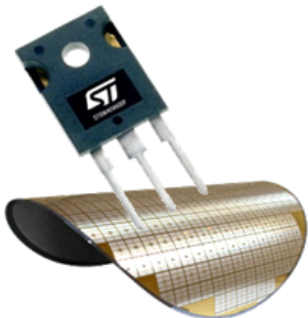
DEVICE	Pin (W)	Input Ic (A)	PF	T [°C]
Competitor 2	2	15.7	0.58	75° - 77°
	3	22.9	0.61	100° - 103°
	~3.8 (MAX)	26.5	7min:30sec	



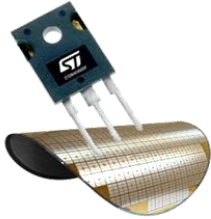
Operating times measured at the maximum input power condition (4kW) just before the welding machine stops running due to the activation of the thermal protection.

full load steady state operation

IGBT P/Ns	$I_{CN}^{1)}$ (A)	$V_{ce(sat)}^{2)}$ (V)	$E_{off}^{3)}$ (mJ)	$t_{sc}^{4)}$ (μ s)	Max T_J ($^{\circ}$ C)	Switching freq. range	FRD Option	Production (MAT 30)
STGW15H120F2	15	2.1	0.4	5	175 $^{\circ}$ C	H (20 - 100kHz)	-	Full production
STGW25H120F2	25	2.1	0.75	5	175 $^{\circ}$ C	H (20 - 100kHz)	-	
STGW40H120F2	40	2.1	1.3	5	175 $^{\circ}$ C	H (20 - 100kHz)	-	
STGW15H120DF2	15	2.1	0.4	5	175 $^{\circ}$ C	H (20 - 100kHz)	Very Fast	WK 17
STGW25H120DF2	25	2.1	0.75	5	175 $^{\circ}$ C	H (20 - 100kHz)	Very Fast	
STGW40H120DF2	40	2.1	1.3	5	175 $^{\circ}$ C	H (20 - 100kHz)	Very Fast	



- 1) I_{CN} : Nominal collector current @ $T_J=100^{\circ}$ C
- 2) $V_{ce(sat)}$: Typical conduction losses @ I_{CN} , $T_J=25^{\circ}$ C
- 3) E_{off} : Typical switching energy losses @ I_{CN} , $T_J=25^{\circ}$ C, $V_{CC}=600$ V
- 4) t_{sc} : min short circuit whitstand time @ $V_{CC}=600$ V, $T=150^{\circ}$ C



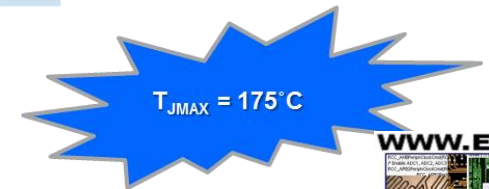
For soft switching applications

Sales Type	$I_{CN}^{1)}$	Main Applications	Production (MAT 30)
STGWT20H125DF	20 A	IH, soft switching	Released
STGWT28IH125DF	30 A	IH, soft switching	



M series (up to 20 kHz)

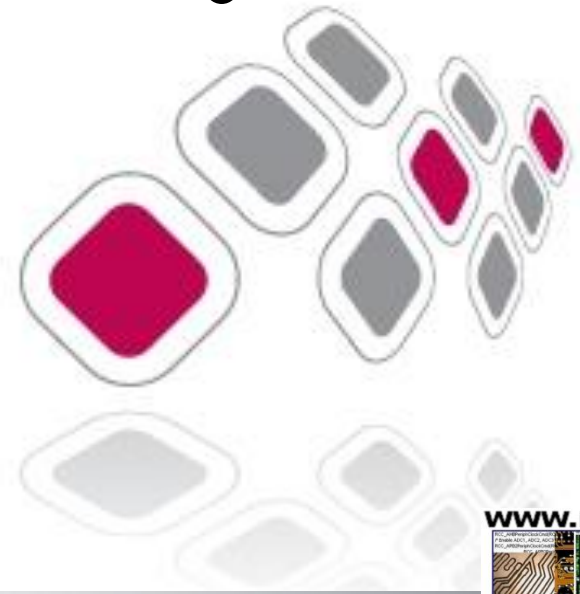
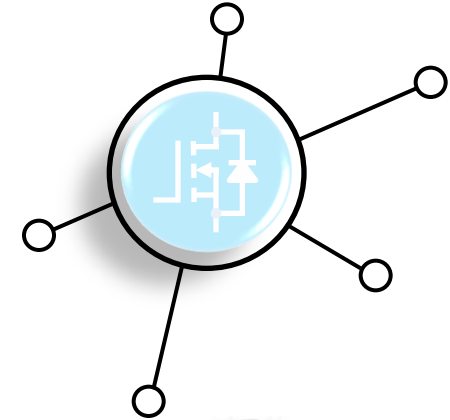
Sales Type	$I_{CN}^{1)}$	$V_{CESAT} @ I_{CN}$	E_{off} (mJ)	Main Applications	Samples	Production (MAT 30)
STGW15M120DF3	15 A	1.85 V	0.9	Drives, UPS, Solar	May '14	wk 24
STGW25M120DF3	25 A	1.85 V	1.5	Drives, UPS, Solar	May '14	wk 24
STGW40M120DF3	40 A	1.85 V	2.35	Drives, UPS, Solar	Available	wk 20



¹⁾ continuous $I_C @ 100^\circ C$

²⁾ Test condition $\rightarrow V_{CC} = 600V, V_{GE} = 15V, T_{Jstart} = 150^\circ C$

Application example with SiC MOSFET and Diodes



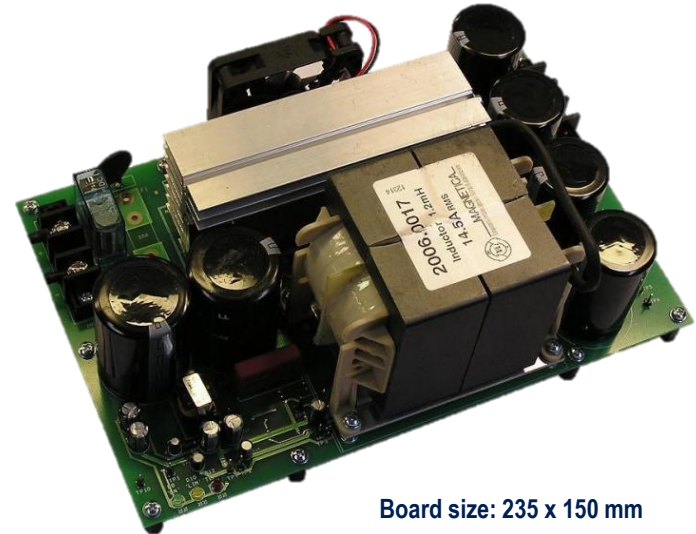
Boost Inverter with SiC semiconductors

- ▶ The goal is to design a 4kW boost inverter to demonstrate the advantages of using ST Silicon Carbide Power MOSFET and Silicon Carbide Schottky rectifiers
- ▶ Design specifications:
 - Input Voltage: 400-600VDC
 - Output Voltage: 800VDC
 - Output Power: 4kW
- ▶ Target efficiency > 99%



Boost Inverter in detail

- Fully integrated and compact solution:
 - Power stage, aux. SMPS, controller, signal processing
- Main ST products:
 - SCT30N120 (1200V / 45A SiC MOSFET)
 - STPSC6H12B (1200V / 6A SiC Diode)
 - TD350ED (GapDrive also tested with equal results)
 - L5991D (current mode PWM controller)
- Optimized for 100kHz switching
- Board available to selected Customers



Board size: 235 x 150 mm

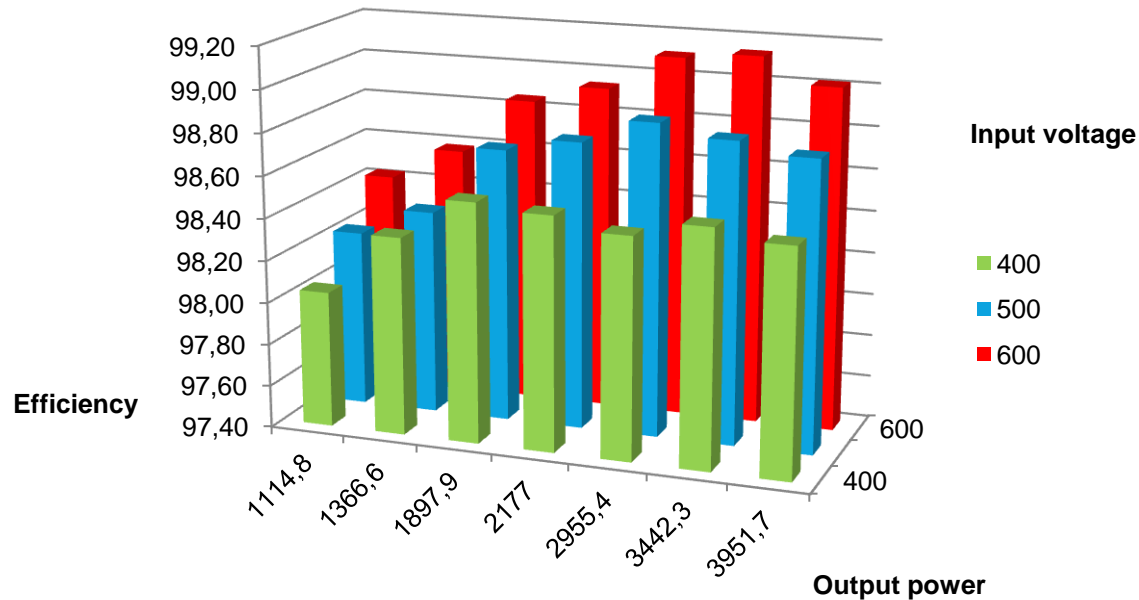
Performance				
Input Voltage (VDC)	Output Power (W)	Heatsink Temperature (°C)	Total efficiency including AUX (%)	Total efficiency without AUX* (%)
600	2094	57.5	99.11	99.29

* Efficiency of boost inverter itself (SiC MOSFET, SiC diodes and main choke)

Testing conditions:

- Input voltage 400, 500 and 600VDC, output voltage 800VDC
- Switching frequency 65kHz, Rgate resistor 2.3OHM
- Output power 1kW to 4kW, Ambient temperature 25°C
- Results: The “peak efficient point” for the demo is around 3.5kW of output load

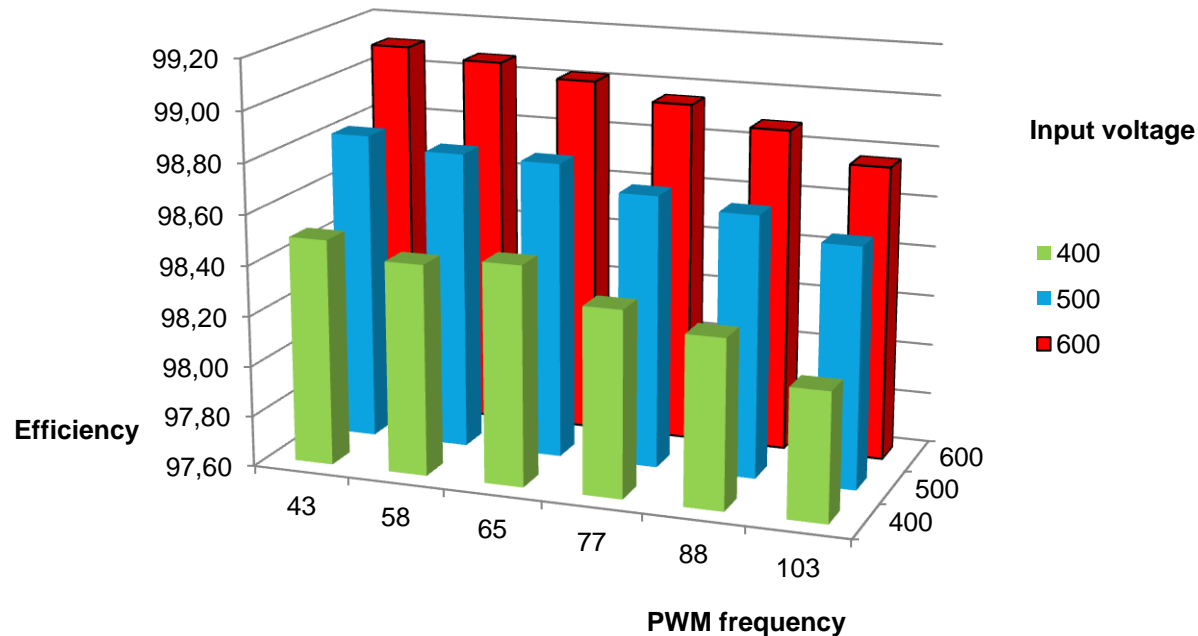
Efficiency versus to output power



Testing conditions:

- ❑ Input voltage 400, 500 and 600VDC, output voltage 800VDC, Rgate resistor 2.3OHM
- ❑ Switching frequency 43kHz, 58kHz, 65kHz, 77kHz, 88kHz, 103kHz
- ❑ Output power - 4kW, Ambient temperature 25°C
- ❑ Results: the overall efficiency of the demo decreasing with increasing of frequency

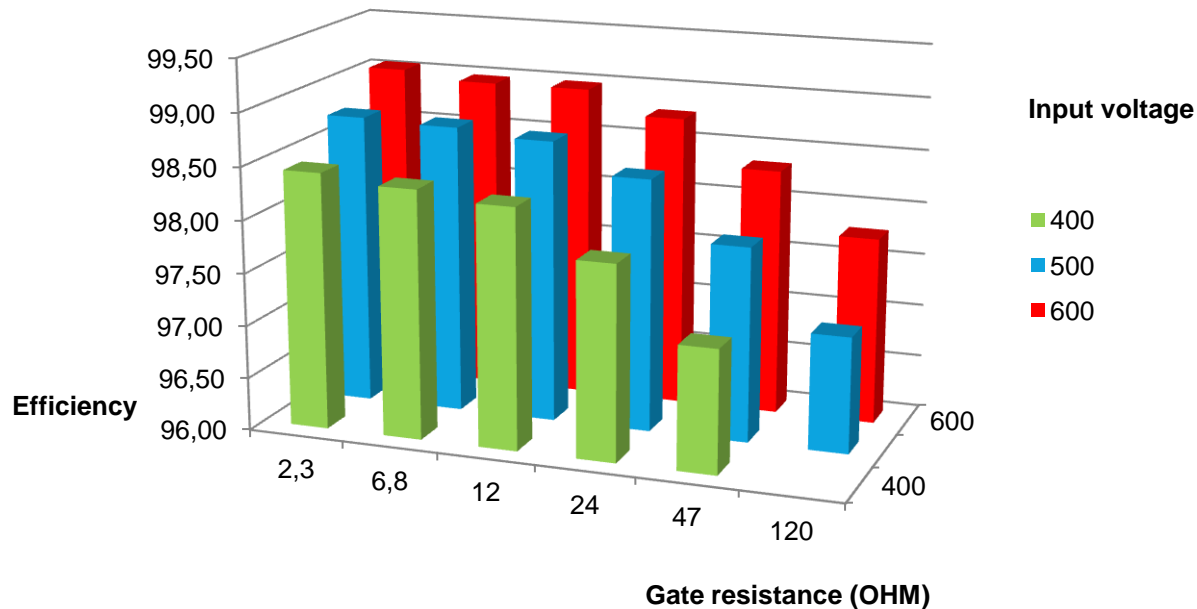
Efficiency versus to PWM frequency



Testing conditions:

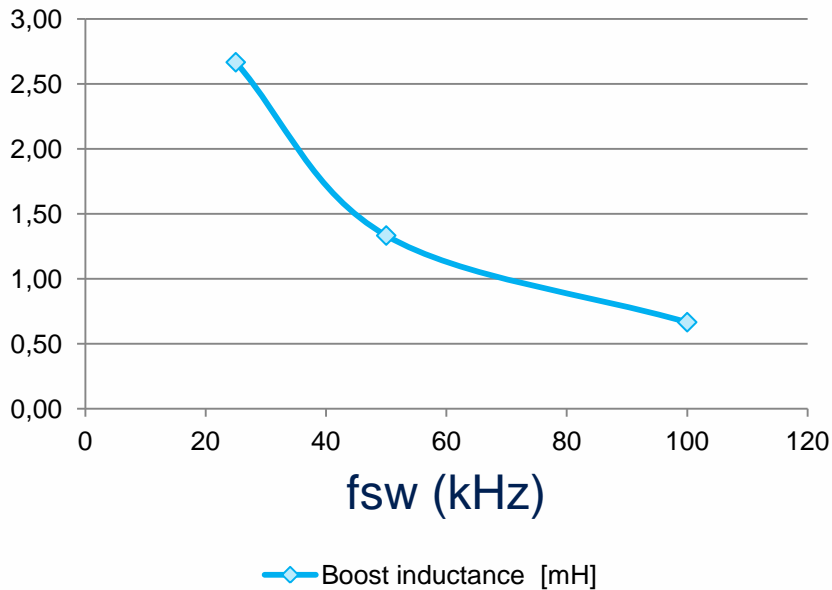
- ❑ Input voltage 400, 500 and 600VDC, output voltage 800VDC, Rgate resistor 2.3OHM
- ❑ Switching frequency 60 kHz
- ❑ Output power - 4kW, Ambient temperature 25°C
- ❑ Results: Increasing of gate resistance decreasing of overall efficiency

Efficiency versus to gate resistance

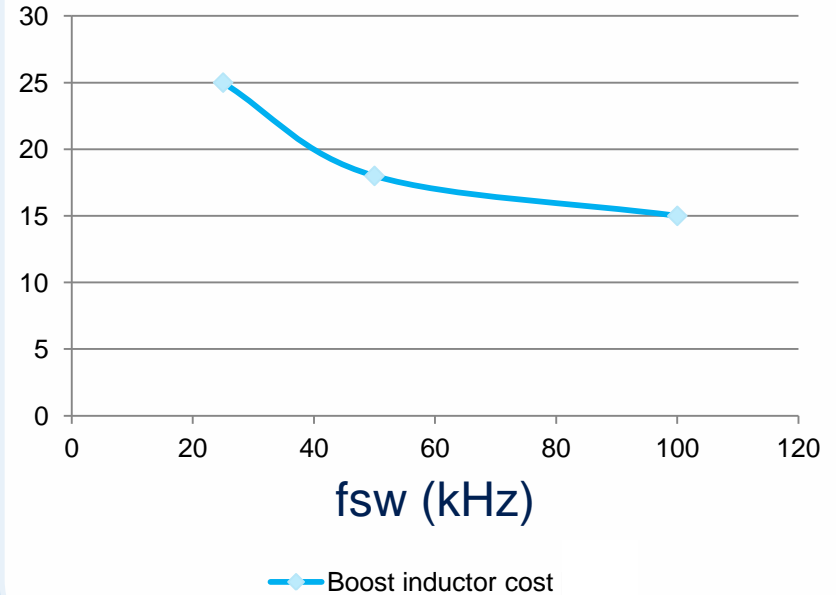


Inductor and cost patten vs. working frequency

Boost inductance (mH)



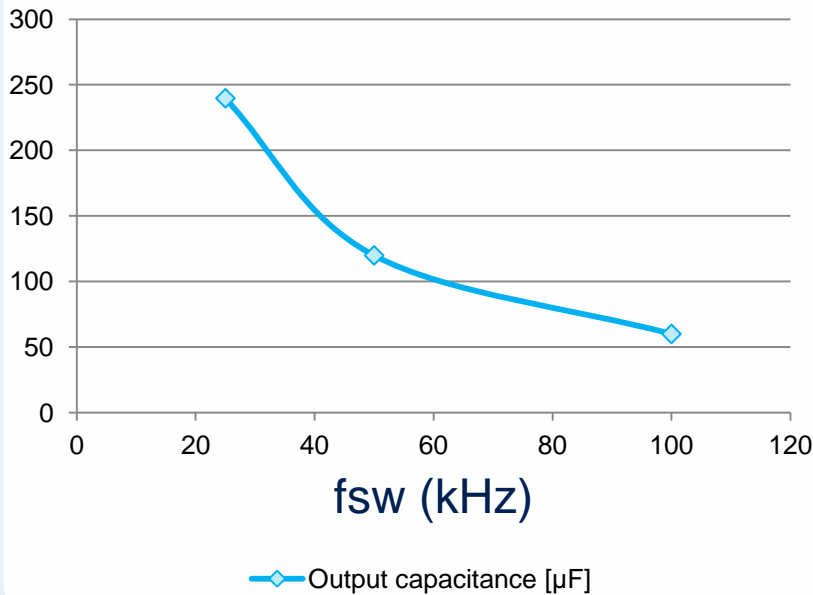
Boost inductor (normalized cost)



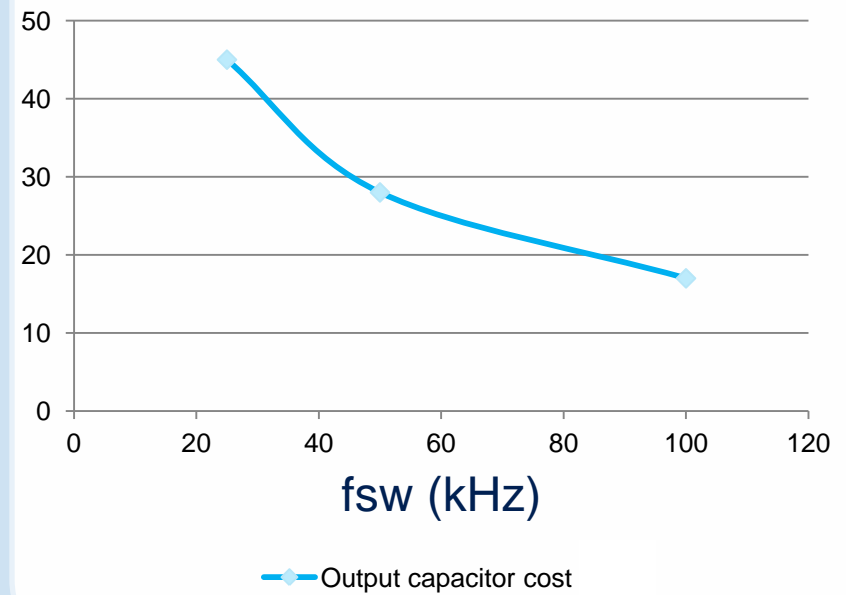
Conditions: 6kW DC/DC Boost Converter in CCM. Worst case for L calculation: $V_{IN}=400V$, $V_{OUT}=800V$, $\Delta I_{RIPPLE}=20\% \cdot I_{AVG-max}$ @25kHz.

Capacitance and cost patter vs. working frequency

Output capacitance (μF)

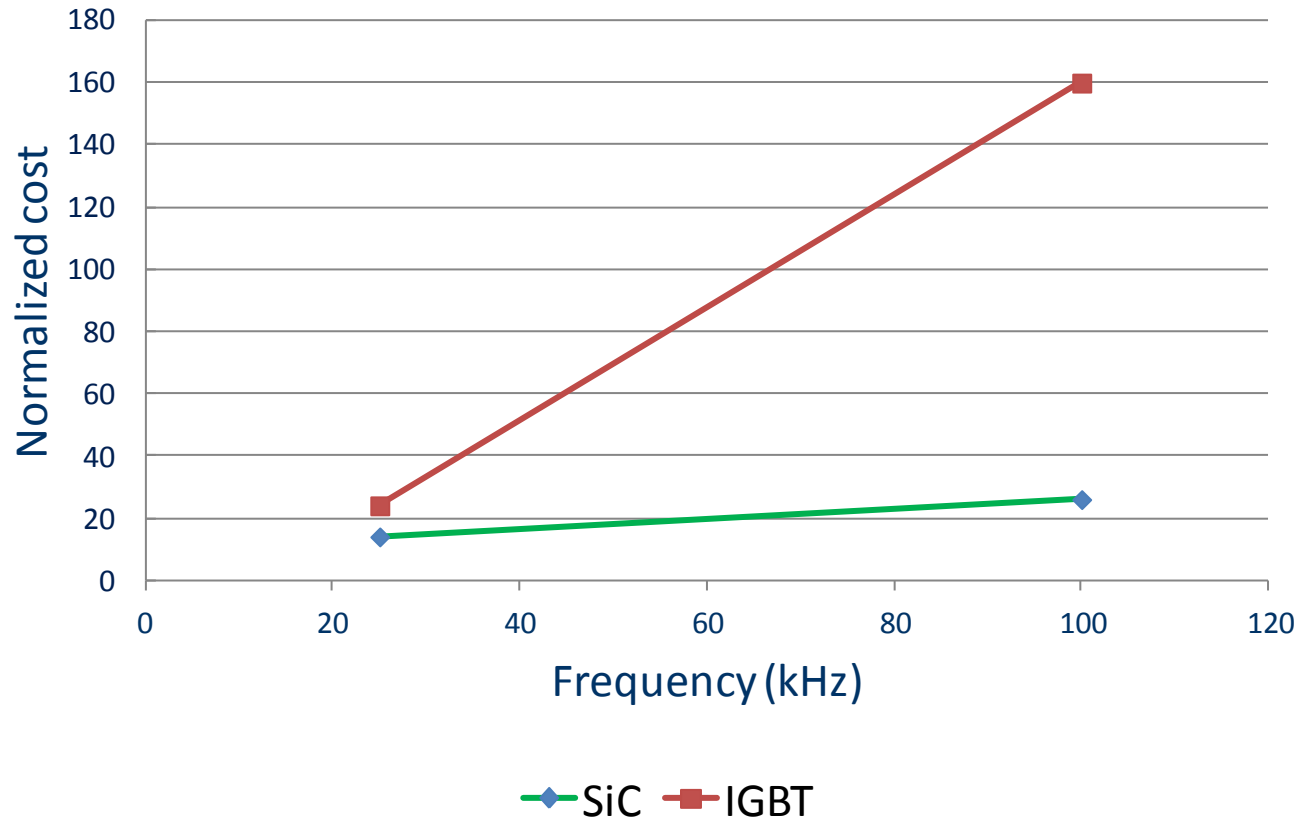


Output capacitor (normalized cost)



Conditions: 6kW DC/DC Boost Converter in CCM. Worst case for C calculation: $V_{IN}=400\text{V}$, $V_{OUT}=800\text{V}$. $\Delta V_{RIPPLE}=1\text{V}$, $L=2.67\text{mH}$ in order to make $\Delta I_{RIPPLE}=20\% \cdot I_{AVG-max}$ @25kHz.

Comparison between 1200V 25A IGBT and SCT30N120 SiC MOSFET



Conditions: DC/DC Boost Converter in CCM, $V_{IN}=600V$, $V_{OUT}=800V$, P_{OUT} up to 6kW, 2xSiC Diodes in parallel as Boost rectifier.



Thanks for your attention

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